

a' How to deal with ~~this~~ this partial dynamics?

Return to structure of  $(C^I) = \begin{pmatrix} C_C^I & C_L^I \\ C_{I,C} & C_{I,L} \end{pmatrix}$  together

with

$$\left[ \begin{array}{l} E_S = \begin{pmatrix} 1 \\ C_S \end{pmatrix} C_C^I \\ \oplus \begin{pmatrix} L_S \\ 1 \end{pmatrix} C_{I,L} \end{array} \right] = \left\{ \begin{pmatrix} V_C & L_S I_L \\ C_S V_C & I_L \end{pmatrix} \right\}$$

You would like to simplify the structure by saying that  $C^I$  ~~has~~ has the structure of a polarized Euclidean space, i.e.  $C^I$  is a Euclidean space equipped with an orthogonal splitting  $C^I = C_C^I \oplus C_L^I$ . This should amount to making the ~~the~~ positive forms  $C_C^I$ ,  $C_L^I$  the identity. ~~From~~ ~~Krein viewpoint~~?

Let  $E$  be a fund. Euclidean space, over  $\mathbb{R}$  pos. symm. form.  $\boxed{H : E \rightarrow E'}$   $H^t = H$

$$|\xi|^2 = \langle \xi | H \xi \rangle > 0 \quad \xi \neq 0$$

Start again  $E$  f.d.  $\mathbb{R}$  vector space,  $E'$  dual space  $H : E \rightarrow E'$  symm. form i.e.

$$\xi_1 \cdot H \xi_2 = \xi_2 \cdot H \xi_1 \quad \boxed{H(\xi_1)(\xi_2) = \xi_1(H\xi_2)}$$

$$H(\xi_2)(\xi_1) = \boxed{\circ}$$

b' Somehow the basic idea is to change variables

$$\tilde{V}_c = C^{1/2} V_c \quad ?$$

$$\tilde{I}_c = C^{-1/2} I_c = C^{-1/2} C s V_c = C^{1/2} s V_c ?$$

$$\begin{pmatrix} C \\ C \end{pmatrix} = \left\{ \begin{pmatrix} V_c \\ I_c \end{pmatrix} \right\} \supset F_1 = \left\{ \begin{pmatrix} V_c \\ C s V_c \end{pmatrix} \right\} = \begin{pmatrix} 1 \\ C \end{pmatrix} C_c^1$$

$$F_1 = \left\{ \begin{pmatrix} V_c \\ C s V_c \end{pmatrix} \right\} = \left\{ \begin{pmatrix} C^{1/2} V_c \\ C^{1/2} V_c \end{pmatrix} \right\} = \left\{ \begin{pmatrix} \tilde{V}_c \\ \tilde{I}_c \end{pmatrix} \right\} ?$$

$$F_1 = \begin{pmatrix} 1 \\ C \end{pmatrix} C_c^1 = \left\{ \begin{pmatrix} V_c \\ I_c = C V_c \end{pmatrix} \right\} = \left\{ \begin{pmatrix} C^{-1/2} V_c \\ C^{1/2} C V_c \end{pmatrix} \right\}$$

$$= \left\{ \begin{pmatrix} C^{-1/2} V_c \\ C^{1/2} V_c \end{pmatrix} \right\} = \left\{ \begin{pmatrix} \tilde{V}_c \\ \tilde{I}_c \end{pmatrix} \right\}$$

still confused

$$\textcircled{1} \quad E \xrightarrow[\sim]{H} E' \quad \boxed{\square} > 0.$$

$$F_H < \begin{pmatrix} E \\ E' \end{pmatrix} \Rightarrow \text{Lagrangian } \cancel{ff} \quad H = {}^t H.$$

c' Sept 26 Focus upon the splitting

$$\begin{pmatrix} C^1 \\ C_1 \end{pmatrix} = \begin{pmatrix} \bar{C}^0 \\ H_1 \end{pmatrix} \oplus \underbrace{\begin{pmatrix} C_s \\ \Gamma_s \end{pmatrix}}_{\parallel} \oplus \underbrace{\begin{pmatrix} L_s \\ \Gamma_{L_s} \end{pmatrix}}_{\parallel}$$
$$\begin{pmatrix} 1 \\ C_s \end{pmatrix} C_c^1 \quad \begin{pmatrix} L_s \\ 1 \end{pmatrix} C_{L_s}$$

for generic s. You expect this to hold for  $\text{Re}(s) \neq 0$ . Find a description of the ~~quotient~~ quotient  $\begin{pmatrix} C^1 \\ C_1 \end{pmatrix} / \Gamma_s$ , actually you want the map  $\begin{pmatrix} \bar{C}^0 \\ H_1 \end{pmatrix} \rightarrow \begin{pmatrix} C^1 \\ C_1 \end{pmatrix} / \Gamma_s$  to be an iso.

You believe that this is equivalent to nondegeneracy of a quadratic form  $Q_s$  depending on ~~s~~ s.

$$\begin{array}{ccc} \bar{C}^0 & \xhookrightarrow{i} & C^1 \longrightarrow H^1 \\ & & \downarrow \bar{Z}_s^{-1} \\ \bullet H_1 & \longrightarrow & C_1 \xrightarrow{i^*} \bar{C}^0 \end{array}$$

This should be obvious.  $Q_s = i^* \bar{Z}_s^{-1} i$ . Get subquotient thm. You need to ~~understand~~ this carefully well on the "Grassmannian" level. Ultimately your objects are real vector spaces with quadratic forms depending on s.

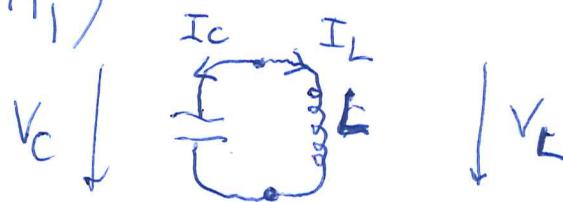
Observe that  $A_s$  on  $\bar{C}^0$  has sing 0,  $\infty$  but on  $H^1$  it has lots of interesting poles, which should give interesting oscillations.

d' ~~Reindeer~~ Aim to ~~P~~

$$\bar{C}^0 \hookrightarrow C^1 \xrightarrow{\beta} H^1$$

$$\bar{C}_0 \leftarrow C_1 \leftrightarrow H_1$$

back to old problem of why there's free motion  
on  $(\bar{C}^0, H_1)$ . ~~This was a state~~



$$\begin{aligned} I_C + I_L &= 0 \\ V_C &= V_L \end{aligned} \quad \} \text{ Kirchhoff}$$

$$\begin{aligned} C_S V_C &= I_C \\ L_S I_L &= V_L \end{aligned}$$

so in terms of dominant variables  $V_C, I_L$  you get

$$C_S V_C = I_C = -I_L = -\frac{1}{L_S} V_L = -\frac{1}{L_S} V_C$$

$$\Rightarrow \left( C_S + \frac{1}{L_S} \right) V_C = 0.$$



from the symplectic picture  
~~approach~~ and quadratic forms

Let's begin with

You want to go  
done with ~~it's necessary~~  
to ~~is~~ Hilbert spaces  
Euclidean

Stop

$$\begin{array}{c} \bar{C}^0 \hookrightarrow C^1 \xrightarrow{\beta} H^1 \\ \downarrow Z_S^{-1} \\ \bar{C}_0 \xleftarrow{\alpha^*} C_1 \leftrightarrow H_1 \end{array}$$

Review the Grass business, i.e. retract of a polarized Hilbert space

$$W \xleftarrow{(\beta_+ \beta_-)} \begin{pmatrix} V_+ \\ V_- \end{pmatrix} \xleftarrow{(\alpha_+ \alpha_-)} W$$

Require  $\beta = \alpha^*$ , i.e.

$$\beta_{\pm} = \alpha_{\mp}^*$$

$$\begin{aligned} \beta_+ \alpha_+ + \beta_- \alpha_- &= 1_W \\ \alpha_+^* \alpha_+ + \alpha_-^* \alpha_- &= 1_W \end{aligned}$$

e'

Retract

$$W \xleftarrow{(\alpha_+^*, \alpha_-^*)} \begin{pmatrix} V_+ \\ V_- \end{pmatrix} \xleftarrow{(\alpha_+, \alpha_-)} W \quad h = \alpha_+^* \alpha_+ + \alpha_-^* \alpha_-$$

$h_+ = \alpha_+^* \alpha_+$  self-adjoint  $0 \leq h_+ \leq 1$ , Suppose  $\alpha_+^* \alpha_+ = 1$   
 $0 < \lambda < 1$ . Then  $\alpha_-^* \alpha_- = 1 - \lambda$ . You find  
 that ???

??

Go back to the Morita equivalence.

~~Retract of~~ Linear retract,  $C$ -module retract  
 of a polarized v.s.  $\begin{pmatrix} V_+ \\ V_- \end{pmatrix}$ . A simple ~~case~~  
 of GNS

Given: the retract  $W \xleftarrow{(\beta_+, \beta_-)} \begin{pmatrix} V_+ \\ V_- \end{pmatrix} \xleftarrow{(\alpha_+, \alpha_-)} W$

$$\frac{h_+}{\beta_+ \alpha_+ + \beta_- \alpha_-} = 1$$

get  $W$  with  $h_+ + h_- = I_W$ . Conversely  
 given  $W$  ——————, can ~~not~~ define

$$W \xleftarrow{\beta_+} V_+ \xleftarrow{\alpha_+} W$$

$$W \xleftarrow{\beta_-} V_- \xleftarrow{\alpha_-} W$$

$(\beta_+, V_+, \alpha_+)$  as canonical factorization of  $h_+$ .

~~GO BACK~~ In the Hilbert space  
 situation you define  $V_\pm$  via completion of  
 $W$  with  $\langle w | h w \rangle$

$f'$

$$W \xrightarrow{\left( \begin{smallmatrix} \lambda^{\frac{1}{2}} & (1-\lambda)^{\frac{1}{2}} \\ 0 & 1 \end{smallmatrix} \right)} \left( \begin{smallmatrix} W & W \\ W & W \end{smallmatrix} \right) \xrightarrow{\left( \begin{smallmatrix} \lambda^{\frac{1}{2}} & 0 \\ 0 & (1-\lambda)^{\frac{1}{2}} \end{smallmatrix} \right)} W$$

so you are in the situation where  $W = V_+ = V_-$   
Not good.

Let's discuss what might happen in general

Recall linear retract of a polarized v.s.

$$\beta \alpha = 1$$

$$W \xleftarrow{\beta = (\beta_+ \beta_-)} \left( \begin{smallmatrix} V_+ & \\ & V_- \end{smallmatrix} \right) \xleftarrow{\alpha = \left( \begin{smallmatrix} \alpha_+ & \\ & \alpha_- \end{smallmatrix} \right)} W$$

$$W \text{ has structure } \mathbb{I}_W = h_+ + h_- \quad h_{\pm} = \beta_{\pm} \alpha_{\pm}$$

Given  $W$  with this partition, then possible ~~if  $\alpha$  is~~ dilations of  $W$  to a polarized v.s. corresponds to factorizations of  $h_+, h_-$ . Hilbert version  $\beta = \alpha^*$ , ~~we have~~ have canon. fact  $V_{\pm}$  = completion of  $W$  wrt  $\langle w | h_{\pm} w \rangle$

Now you are concerned (with) not only ~~with~~  $W$  but also with  $W^\perp = \text{Ker}(\beta) = \text{Coker}(\alpha)$ . How to make progress?

Idea:  $\Gamma_s \subset \left( \begin{smallmatrix} C^1 \\ C_1 \end{smallmatrix} \right)$  is a correspondence

from  $C^1$  to  $C_1$  defined for all  $s \in P_1$ ?

Why? Split into  $L, C$  summands. ~~Then~~ you have

$$\Gamma_{L,s} = \left( \begin{smallmatrix} sL \\ 1 \end{smallmatrix} \right) C_1 \quad L \text{ invertible}$$

$$\Gamma_{C,s} = \left( \begin{smallmatrix} 1 \\ sc \end{smallmatrix} \right) C^1 \quad C \quad "$$

g' Problems, Questions, You want to review  
~~Homework~~ all you know about  
 LC networks.

An LC network is a connected graph such that each edge is either a capacitor (having capacitance  $\in (0, \infty)$ ) or an inductor (having inductance  $\in (0, \infty)$ ).

The kinematics of such a network depends only on the underlying graph. The <sup>vector</sup> space of ~~states~~ <sup>current</sup> states spanned of the network is the direct sum

$$\begin{pmatrix} C^1 \\ C_1 \end{pmatrix}$$

$$\begin{pmatrix} V \\ I \end{pmatrix}$$

of the space of Real valued 1-cochains and 1-chains. Aug

elt.  $V$  of  $C^1$  is a family of voltage drops indexed by the edges; an elt  $I$  of  $C_1$  is a family of currents indexed by the edges. The natural pairing between 1-cochains and 1-chains yields the power of the state  $\begin{pmatrix} V \\ I \end{pmatrix} \in \begin{pmatrix} C^1 \\ C_1 \end{pmatrix}$

$$\text{power} = \sum_{\text{edges } e} V_e I_e = V \cdot I$$

~~that's what I do~~

The ~~state of the~~ state space has a symplectic bilinear form given by

$$\omega \left( \begin{pmatrix} V \\ I \end{pmatrix}, \begin{pmatrix} V' \\ I' \end{pmatrix} \right) = V \cdot I' - I \cdot V'$$

(Also) a symmetric form  $V \cdot I' + I \cdot V'$  equivalent to power as a quad. form

so far have been discussing the unconstrained state spaces associated ~~to~~<sup>the</sup> the edges of the network.

~~Now introduce constraints given by Kirchhoff's laws:~~

2 laws:  $V$  is 1-coboundary i.e.  $V = \delta\phi$ ,  $\phi \in \bar{C}^0$   
 $I$  is 1-cycle i.e.  $\partial I = 0$

Get short ~~exact sequence + its dual:~~

$$\bar{C}^0 \hookrightarrow C^1 \twoheadrightarrow H^1$$

$$\bar{C}_0 \leftarrow C_1 \leftarrow H_1$$

Can also write this

$$\begin{pmatrix} \bar{C}^0 \\ H_1 \end{pmatrix} \hookrightarrow \begin{pmatrix} C^1 \\ C_1 \end{pmatrix} \twoheadrightarrow \begin{pmatrix} H^1 \\ \bar{C}_0 \end{pmatrix}$$

↴  
 Lag subspace  
 $W$ .  
 ↴  
 symplectic  
 v. space  
 ↴  
 dual space  
 $W^*$

to emphasize the symplectic structure. It's not clear that this is useful.

Dynamics: ~~Consider first case~~

For a capacitor  $C \frac{1}{F}$  a time-dependent state

$(V(t) \quad I(t))$  satisfies the DE  $C\dot{V} = I$ , for  $L \frac{1}{F}$ , a time dep state satisfies the DE  $L\dot{I} = V$ .

Use L.T. to time dependent or frequency dependent states.

You have  $\frac{d}{dt} \mapsto$  mult by s under the L.T.

so you get  $\begin{cases} CsV = I & \text{for capacitor edge} \\ LsI = V & \text{for inductor edge} \end{cases}$

Now you can describe<sup>a</sup> free motion of the LC circuit, namely, it's a state  $\begin{pmatrix} V(t) \\ I(t) \end{pmatrix}$  depending on t whose values in the Kirchhoff space, in other words a path in the Kirchhoff space. Also it should satisfy the dynamical conditions  $C\dot{V} = I$ ,  $L\dot{I} = V$  for C (resp. L) edges.

Better to look first for exponential solutions  $e^{st} \begin{pmatrix} \tilde{V} \\ \tilde{I} \end{pmatrix}$ , and later worry about completeness.

Actually it shouldn't be hard. If  $\begin{pmatrix} V(t) \\ I(t) \end{pmatrix}$  is a solution, then so is  $\begin{pmatrix} V(t-t_0) \\ I(t-t_0) \end{pmatrix}$  ??

Assuming the path  $\begin{pmatrix} V(t) \\ I(t) \end{pmatrix}$  is differentiable  $C^\infty$  in t, then you can differentiate to get others. ??

Recap. ~~What do we know?~~ The

basic object to study is an exponential solution  
 $e^{\text{st}}(V)$  satisfying the 2 Kirchhoff relations  
and the dynamical relations:

$$V \in \bar{C}^\circ$$

$$\cancel{Cs} V_C = I_C \quad \left(\frac{V_C}{I_C}\right) e\left(\begin{matrix} C_C^1 \\ C_{IC} \end{matrix}\right)$$

$$I \in H_1$$

$$Ls I_L = V_L \quad \left(\frac{V_L}{I_L}\right) e\left(\begin{matrix} C_L^1 \\ C_{IL} \end{matrix}\right)$$

Draw this

$$\bar{C}^\circ \hookrightarrow C_C^1 \oplus C_L^1 \longrightarrow H^1$$

$$\left(\begin{smallmatrix} 1 \\ Ls \end{smallmatrix}\right) C_C^1 \oplus \left(\begin{smallmatrix} Ls \\ 1 \end{smallmatrix}\right) C_{IL} = \Gamma_s$$

$$\bar{C}_0 \leftarrow C_{IC} \oplus C_{IL} \longrightarrow H_1$$

Meaning: An exp. soln  $e^{\text{st}}(V)$  amounts to  
a  $V \in \bar{C}^\circ$ ,  $I \in \cancel{H}_1$  such that the image of  
 $V$  in  $C^1$  and  $\bar{C}^\circ \xrightarrow{\quad} C^1 \longrightarrow H^1$  the image of  
 $I$  in  $H_1$  are  
related by an  
elt of  $\Gamma_s$

$$\bar{C}_0 \longleftrightarrow C^1 \longleftrightarrow H_1$$

At this point you understand exp. solutions  
in terms of the symplectic picture. This is where  
you have been stuck for a long time.

The way out of the confusion is to ~~restrict to~~ restrict to  
the voltage (cochain) side, since the ~~current~~ current (chain)  
side can be reconstructed by duality.  $\Gamma_s$  is  
equivalent to a quad form

k' Things you need to know

What kind of resolvent you get when  
 $\Gamma_s \subset \binom{C'}{C_1}$  is transversal to  $\binom{\bar{C}^0}{H_1}$

You need to check this condition is equivalent to the restriction of the energy quadratic form is nondegenerate on  $\bar{C}^0$

Review. Given a connected LC network, a graph whose edges are either capacitors or inductors. Short exact sequence

$$\begin{array}{ccccc} \bar{C}^0 & \longrightarrow & C^1 & \longrightarrow & H^1 \\ & & \parallel & & \\ & & \binom{C'_C}{C'_L} & \ni & \binom{V_C}{V_L} \\ & & & & t V_C \circ V_C \quad t V_L \circ V_L \end{array}$$

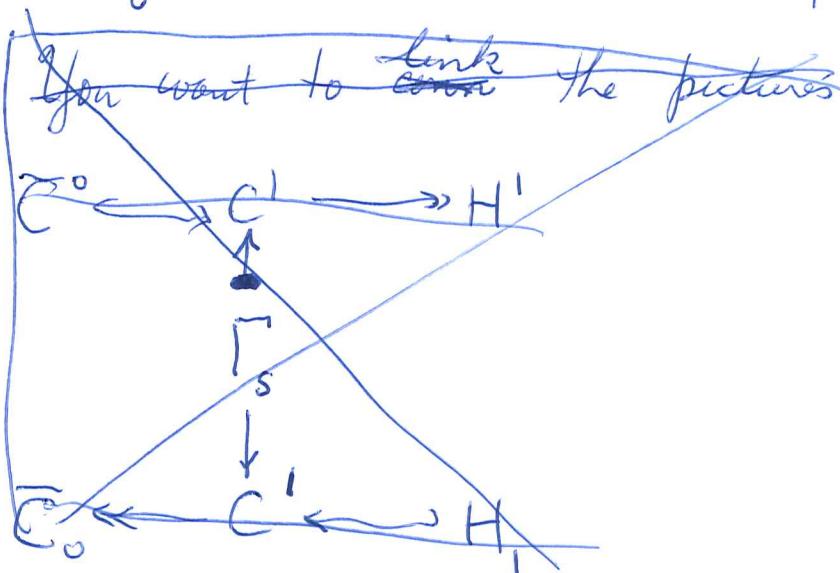
have quadratic form (s dep)

$$Q_s(V) = s \boxed{CV_C^2} + s^{-1} \boxed{LV_L^2}$$

$Q_s$  should be equivalent to the correspondence  $\Gamma_s \subset \binom{C'}{C_1}$

$$\Gamma_{C,s} = \binom{1}{sC} C_C^1, \quad \Gamma_{L,s} = \binom{sL}{1} C_L^1$$

Note: this family of correspondences  $\Gamma_s$  is a ~~vector~~ subbundle over  $P^1$  of the trivial bundle with fibre  $\binom{C'}{C_1}$ .



$\ell'$  You have  $\mathbb{Q}$  on  $C'$  which ~~splits~~  
~~gives~~ an ~~is~~ for generically,  
splits the short exact sequence  $\bar{C}^0 \hookrightarrow C' \rightarrow H^1$ ,  
orthogonally + yields quad form on  $\bar{C}^0 \hookrightarrow H^1$ .

go over  
equivalence

$$\begin{array}{c} \bar{C}^0 \hookrightarrow C' \twoheadrightarrow H^1 \\ \downarrow \Gamma_s \\ \bar{C}_0 \leftarrow C_1 \leftarrow H_1 \end{array}$$

④  $\left( \begin{matrix} \bar{C}^0 \\ H_1 \end{matrix} \right) \cap \Gamma_s = 0 \Leftrightarrow \forall V \neq 0 \in C^1 \langle V, \Gamma_s V \rangle \neq 0$

You need to calculate

$$\left( \begin{matrix} C^1 \\ C_1 \end{matrix} \right) / \Gamma_s \stackrel{?}{=} C^1 \text{ equipped with power quadratic form}$$

Review a little

You believe that  $\left( \begin{matrix} \bar{C}^0 \\ H_1 \end{matrix} \right) \cap \Gamma_s$  is the space of

exponential solutions of the linear equations  
with time behavior  $e^{st}$ . This seems obvious.

$$\begin{aligned} \text{Sing} &= \left\{ s \in \mathbb{P}^1 \mid \left( \begin{matrix} \bar{C}^0 \\ H_1 \end{matrix} \right) \cap \Gamma_s \neq 0 \right\} \\ \text{Reg} &= \left\{ s \in \mathbb{P}^1 \mid \Gamma_s = 0 \right\} \end{aligned}$$

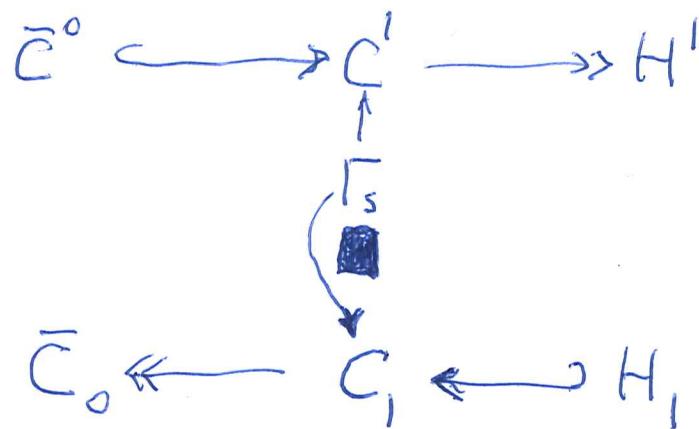
④ For  $s \in \text{reg}$  you get a splitting

$$\left( \begin{matrix} C^0 \\ C_1 \end{matrix} \right) = \left( \begin{matrix} \bar{C}^0 \\ H_1 \end{matrix} \right) \oplus \Gamma_s$$

which ~~should~~ might mean that you get an induced

$m'$  quadratic form on  $\begin{pmatrix} \bar{C}^0 \\ H_1 \end{pmatrix}$  @ 00

conn LC network



$$\begin{pmatrix} \bar{C}^0 \\ H_1 \end{pmatrix} \subset \begin{pmatrix} C' \\ C_1 \end{pmatrix} \Rightarrow \Gamma_s$$

You believe  $\begin{pmatrix} \bar{C}^0 \\ H_1 \end{pmatrix} \cap \Gamma_s$  is the space of

exponential solutions with time dependence  $e^{st}$   
if  $\begin{pmatrix} \bar{C}^0 \\ H_1 \end{pmatrix} \cap \Gamma_s = 0$  in which  
case one has an eigen

$$\begin{pmatrix} \bar{C}^0 \\ H_1 \end{pmatrix} \xrightarrow{\sim} \begin{pmatrix} C' \\ C_1 \end{pmatrix} / \Gamma_s$$

Natural question

Given  $A: V \rightarrow V^*$  symmetric, what is

$$\frac{(V)}{(V^*)} / \frac{(I)}{(A)} V$$

$$V^* \xleftarrow{(-A \ I)} \begin{pmatrix} V \\ V^* \end{pmatrix} \xleftarrow{(I \ A)} V$$

$$\begin{array}{ccc} & V & \\ \xrightarrow{(I \ A)} & \begin{pmatrix} V \\ V^* \end{pmatrix} & \xrightarrow{(-A \ I)} V^* \\ \downarrow (I \ 0) & & \downarrow (-A) \\ V & & V^* \end{array}$$

It looks like  $\begin{pmatrix} V \\ V^* \end{pmatrix} / \Gamma_A \xrightarrow{\sim} V^*$

better would be

$$\begin{array}{ccc} V & \xrightarrow{(I \ 0)} & \begin{pmatrix} V \\ V^* \end{pmatrix} \\ & & \xrightarrow{} \begin{pmatrix} V \\ V^* \end{pmatrix} / \Gamma_A \\ \begin{pmatrix} C^0 \\ H_1 \end{pmatrix} & \xrightarrow{} & \begin{pmatrix} C^1 \\ C_1 \end{pmatrix} / \Gamma_S \end{array}$$

what you want to know is that there is  
a commutative diag

$$\begin{array}{ccc} V & \xrightarrow{\sim} & \begin{pmatrix} V \\ V^* \end{pmatrix} / \Gamma_A \\ A \downarrow & & \downarrow \\ V^* & \xrightarrow{} & \end{array}$$

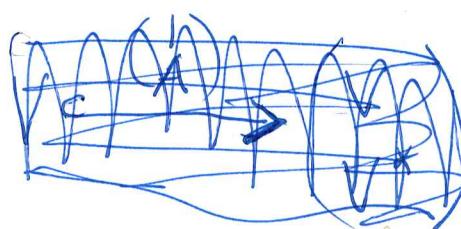
Given  $A: V \rightarrow V^*$  symm.

To understand  $(V) / \Gamma_A$ .

$$V \xrightarrow{(A)} (V) / \Gamma_A$$

$$(0, 1) \downarrow$$

$$V^*$$



$$V \xleftarrow{(A)} (V) / \Gamma_A$$

but:  $(0, 0)$   
 $(0 A^{-1})$

no obvious complement

$$(C^\circ)_{H_1} \xrightarrow{\text{in Reg case}} (C^l) / \Gamma_A \xrightarrow{(-A, 1)} C_l$$

$$\text{You get then an isom } (C^\circ)_{H_1} \xrightarrow{\sim} C_l$$

$$\text{given by } \begin{pmatrix} V^\circ \\ I^l \end{pmatrix} \mapsto -AV^\circ + I^l.$$

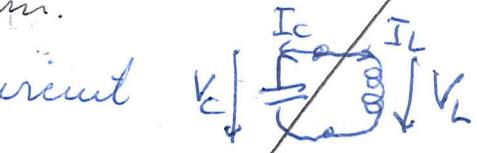
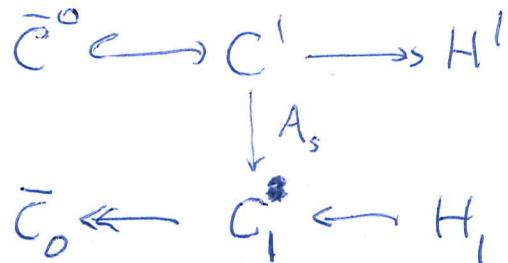
words

$$\begin{array}{ccccc} C^\circ & \hookrightarrow & C^l & \rightarrow & H^l \\ \downarrow S & & \downarrow A & & \\ E^\circ_0 & \leftarrow & C_l & \leftarrow & H_l \end{array}$$

Also have  $A^{-1}$   
 $C^l$   
 $\downarrow$   
 $\text{so you get an } C_l$   
 $\text{induced quad form}$   
 $\text{on } H_l$

You still don't understand the motion of an LC network. You want to derive, construct it from the singularities of a frequency dependent quadratic form.

So let's study the simple circuit



$$\{V\} \quad \{V_{C,L}\}$$

$$I_o \quad \{I_C, I_L\} \quad I_x$$

Notation

$V_o$	$V_C$	$V_L$	$V_x$
$I_o$	$I_C$	$I_L$	$I_x$

~~$\{V_{C,L}\}$~~

$V_o \mapsto (V_o, V_o)$

$V_C \quad V_L$

$V_C, V_L \mapsto V_x = V_C - V_L$

$\begin{matrix} & V_C \\ \uparrow & \downarrow \\ C_S V_C & L_S V_L \\ \uparrow & \downarrow \\ I_C & I_L \end{matrix}$

$F_S = \left(\begin{matrix} 1 \\ A_s \end{matrix}\right) C'$

$I_o = I_C + I_L \leftarrow (I_C, I_L)$

~~$(V | I_C) = (V | I_L) = (V | I_o) = (V | i^*(I_C))$~~

$(V | I_C) = (V | I_L) = (V | I_o) = (V | i^*(I_C))$

$(I_x | V_C) = (I_x | V_L) = (I_x | V_C - V_L)$

$$\begin{array}{c}
 \text{Diagram showing state-space representation: } \\
 \bar{C}^0 \xrightarrow{C} C^1 \xrightarrow{f} H^1 \\
 \text{with } A_s \text{ and } f^* \text{ indicated.} \\
 \text{Equations:} \\
 A_s \left( \begin{matrix} V_C \\ V_L \end{matrix} \right) = \left( \begin{matrix} C_s V_C \\ (L_s)^{-1} V_L \end{matrix} \right) \in C_1 \\
 j(V_C) = V_C - V_L \\
 f^*(I) = \left( \begin{matrix} I_C \\ -I_L \end{matrix} \right) \\
 f^*(I_C) = I_C + I_L \\
 j(I_\lambda) = \left( \begin{matrix} I_\lambda \\ -I_\lambda \end{matrix} \right)
 \end{array}$$

$$\begin{array}{c}
 \text{Diagram showing state-space representation: } \\
 \bar{C}^0 \xrightarrow{C} C^1 \xrightarrow{f^*} H^1 \\
 \text{with } A_s \text{ and } f \text{ indicated.} \\
 \text{Equations:} \\
 \{V_o \in \mathbb{C}\} \quad A_s \left( \begin{matrix} \{V_C\} \in \mathbb{C}^2 \end{matrix} \right) \quad \{V_\lambda \in \mathbb{C}\} \quad j^*(V_C) = V_C - V_L \\
 \{I_o \in \mathbb{C}\} \quad \left\{ \begin{matrix} \{I_C\} \in \mathbb{C}^2 \end{matrix} \right\} \quad \{I_\lambda \in \mathbb{C}\} \quad j^*(I_C) = I_C + I_L \\
 j(I_\lambda) = \left( \begin{matrix} I_\lambda \\ -I_\lambda \end{matrix} \right)
 \end{array}$$

$$(V_o \mid \left( \begin{matrix} I_C \\ I_L \end{matrix} \right)) = (V_o \mid i^*(I_C)) = (V_o \mid I_C + I_L) = (V_o \mid i^*(I_C))$$

$$(jI_\lambda \mid \left( \begin{matrix} V_C \\ V_L \end{matrix} \right)) = ((I_\lambda) \mid \left( \begin{matrix} V_C \\ V_L \end{matrix} \right)) = (I_\lambda \mid V_C - V_L) = (I_\lambda \mid j^*(V_C))$$

$$A_s \left( \begin{matrix} V_C \\ V_L \end{matrix} \right) = \left( \begin{matrix} C_s V_C \\ (L_s)^{-1} V_L \end{matrix} \right) \in C_1$$

Next you want to understand the splitting for  $s$  regular. ~~It's not always good~~

$$\begin{array}{c} (\bar{C}^0) \\ \downarrow H_1 \\ \text{---} \\ (\bar{C}^1) \\ \leftarrow \Gamma_s \end{array}$$

Your aim is to ~~prove~~ for generic  $s$  that  $(\bar{C}^1)$  is the direct sum of the subspaces  $(\bar{C}^0)$  and  $\Gamma_s$ . Viewpoint? First idea: Have short exact seq

$$\begin{array}{ccccc} (\bar{C}^0) & \hookrightarrow & (\bar{C}^1) & \twoheadrightarrow & (H^1) \\ \downarrow \Gamma_s & & \downarrow & & \nearrow \text{using?} \\ \text{and subspace of} & & & & \end{array}$$

Second idea: You have a short exact sequence

$$\bar{C}^0 \hookrightarrow C^1 \twoheadrightarrow H^1$$

and a quadratic form  $A_s$  on  $C^1$ . Suppose  $s$  real  $> 0$ . Then  $A_s$  is positive definite on  $C^1$ , there's a canonical splitting as a Euclidean space.

$s'$   $C^1$  is a Euclidean space,  $\bar{C}^0$  splits orthogonally into  $\bar{C}^0$  and its orthogonal complement, which is canon isom. to  $H^1$ :

$$\bar{C}^0 \xrightarrow{\iota} C^1 \xleftarrow{\pi} (\bar{C}^0)^\perp \cong H^1$$

$$\begin{array}{ccc} & (\bar{C}^0)^\perp & \\ & \downarrow \pi & \swarrow \\ \bar{C}^0 & \xrightarrow{\iota} & C^1 \xrightarrow{\quad} H^1 \end{array}$$

$$\begin{array}{ccccc} & \bar{C}_0 & \xleftarrow{\iota^*} & C_1 & \xleftarrow{\quad} H_1 \\ & \swarrow & & \downarrow j^* & \swarrow \\ & & & H_1^\perp & \end{array}$$

You are still trying to achieve the splitting of  $(C^1)$  into  $(\bar{C}^0)$  and  $\begin{pmatrix} 1 \\ A_s \end{pmatrix} C^1 = \begin{pmatrix} A_s^{-1} \\ 1 \end{pmatrix} C_1$ .

Not really, because you are exploring the 2nd idea: Exact seq  $\bar{C}^0 \hookrightarrow C^1 \rightarrow H^1$  + quadratic form  $A_s$  on  $C^1$ , which is pos. for  $s > 0$ . You know the exact seq splits canonically, and further that  $\bar{C}^0, H^1$  inherit positive quad. forms compatible with the splitting.

Where do you get a Green's function from all this?

$$V_o \xrightarrow{t^*} \begin{pmatrix} V_o \\ V_c \\ V_L \end{pmatrix} \downarrow A_s$$

$$\left( C_s + \frac{1}{Ls} \right) V_o \xleftarrow{t^*} \begin{pmatrix} C_s V_o \\ (Ls)^{-1} V_o \end{pmatrix}$$

$$\begin{pmatrix} V_c = \frac{1}{C_s} I_x \\ V_L = -Ls I_x \end{pmatrix} \xrightarrow{t^*} \left( \frac{1}{C_s} + Ls \right) I_x = V_L$$

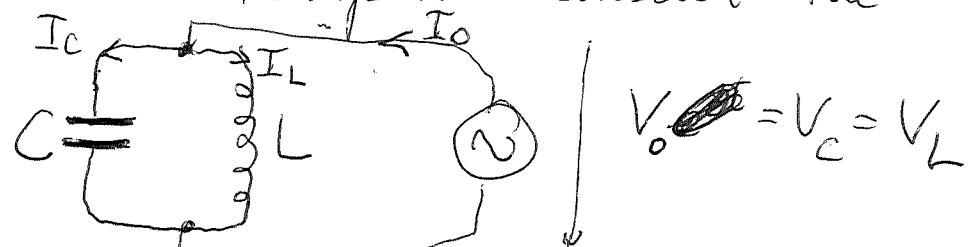
$$\uparrow A_s^{-1}$$

$$\begin{pmatrix} I_c = I_x \\ I_L = -I_x \end{pmatrix} \xleftarrow{t^*} I_x$$

~~exact seg + pos~~ Review the situation. ~~Consider~~ Consider  
quad form  $A_s$  ( $s > 0$ )

$$\bar{C}^0 \xrightarrow{t^*} C^1 \longrightarrow H_1$$

You want a suitable Green's fn. ~~What~~ What  
do you mean? ~~Example.~~ Consider the  
simple oscillator



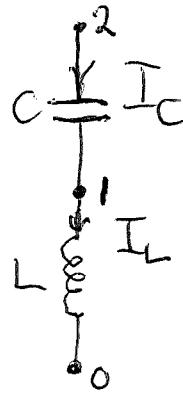
and apply emf of frequency  $s$ .

$$I_c = C_s V_o, \quad I_L = \frac{1}{Ls} V_o$$

$$I_o = I_c + I_L$$

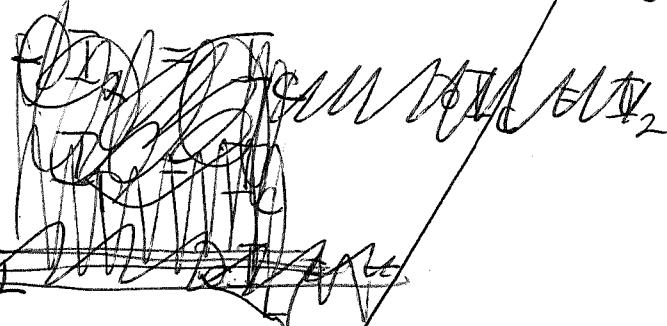
$$I_o = \left( C_s + \frac{1}{Ls} \right) V_o$$

$u'$



$$\bar{C}^0 \cong C^1 = \left\{ \begin{pmatrix} V_C \\ V_L \end{pmatrix} \right\}$$

$$\bar{C}_0 \cong C_1 = \left\{ \begin{pmatrix} I_C \\ I_L \end{pmatrix} \right\}$$



$$\partial[C] = [2] - [1]$$

$$\partial[L] = [1] - [0]$$

~~REVIEW~~ Yesterdays lesson: the free oscillations, normal modes result from the singularities introduced by pushing ~~restriction~~ the quadratic form  $A_s$  down on  $C^1$  down to  $H^1$ .

Review formulas.

$$0 \rightarrow X \xrightarrow{\iota} Y \xrightarrow{f} Z \rightarrow 0$$

~~REVIEW~~  $y^t A y$  quad form on  $Y$

$x^t \iota^t A \iota x$  restriction of  $y^t A y$  to  $X$ . Assume  $\iota^t A \iota$  is non degenerate i.e. ~~there~~  $X \xrightarrow{\sim} X^*$ . Then there ~~should be~~ a push forward  $f_* A$  <sup>quad form</sup> ~~defined~~ whose value at  $z$  is the stationary value of  $A$  on the ~~base~~ <sup>affine space</sup>  $f^{-1} z$ .

Pick  $y_0 \in f^{-1} z$   $f^{-1} z = \{y_0 + cx \mid x \in X\}$

$$(y_0 + cx)^t A (y_0 + cx) = y_0^t A y_0 + x^t \iota^t A y_0 + y_0^t A c x + \dots + x^t \iota^t A c x$$

$$0 \rightarrow X \rightarrow Y \rightarrow Z \rightarrow 0$$

$$\text{of variation } S_y = y + i\delta x$$

$$\cancel{S(y^t A y) = (y + i\delta x)^t A y + y^t A (y + i\delta x)}$$

if  $y$  is a stationary point

$$\cancel{S(y^t A y) = (y + i\delta x)^t A y + y^t A (y + i\delta x)}$$

Let  $y_0$

$$y + \delta y = \cancel{y} y + i\delta x$$

$$S(y^t A y) = y^t A i\delta x + \cancel{(i\delta x)^t A y} = 2 \delta x^t i^t A y$$

Assume zero for all  $\delta x \Rightarrow i^t A y = 0$

says  $y$  is  $\perp$  to  $iX$ .

~~the suppose~~  $y = y_0 + i x$

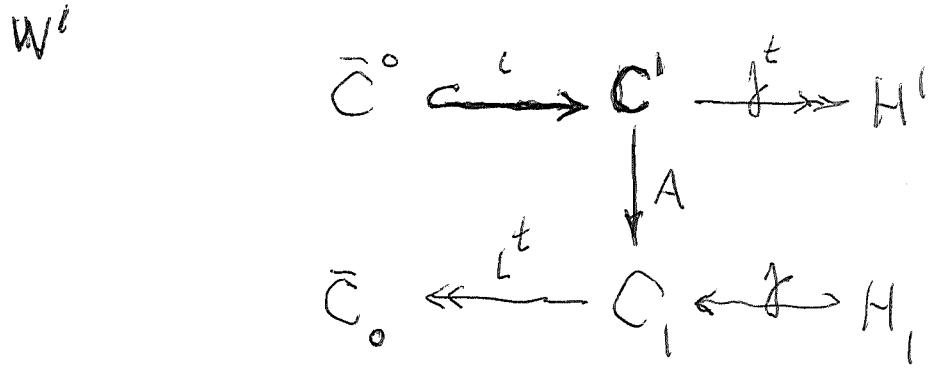
$$i^t A y_0 + i^t A i x = 0$$

$$\Rightarrow x = -(i^t A i)^{-1} i^t A y_0$$

so the ~~critical~~ point ~~is~~ is

$$y = y_0 - i(i^t A i)^{-1} i^t A y_0$$

$$i^t A y = 0$$

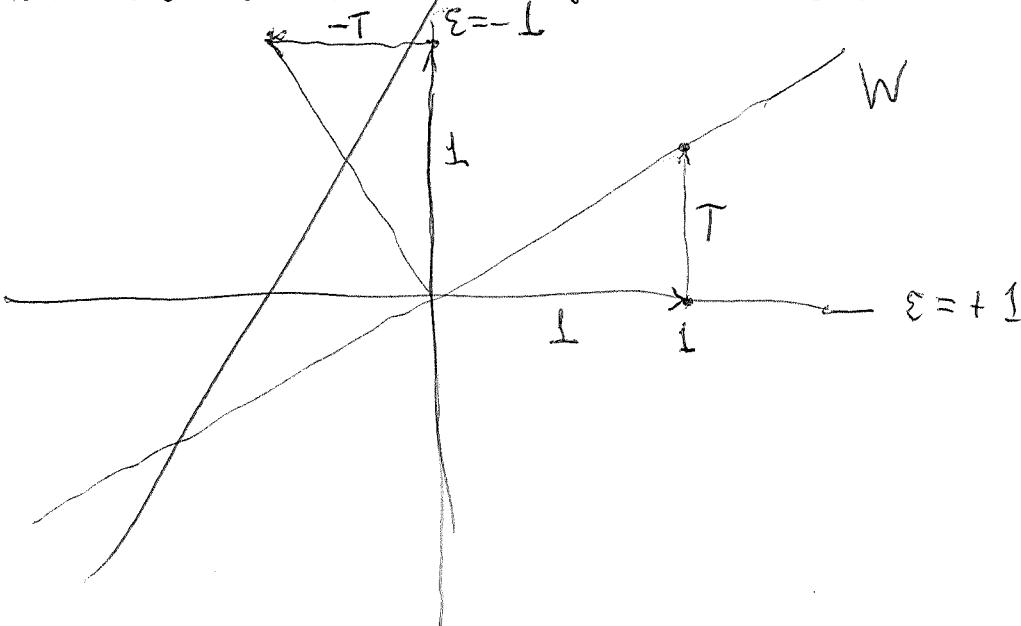


$$y \in C^1 \text{ sat } \iota^t A y = 0 \quad \text{means} \quad \begin{aligned} & Ay = f y \\ & y = A^{-1} f y \\ & f^t y = (f^t A^{-1} f) y \end{aligned}$$

$$W \xrightarrow{\iota} \begin{pmatrix} V^+ \\ V^- \end{pmatrix} \xrightarrow{f^t} W$$

Representation of dihedral group gen by  $F, \varepsilon$   
 $g = F\varepsilon ?$  Put into words what you want.

decompose  
 You want to ~~find~~ the rep of  $\langle F, \varepsilon \rangle$  into irreducibles. What are the irreducibles?



$x'$  Grassmannian details.

$$W \xleftarrow{\iota^*} \begin{pmatrix} V_+ \\ V_- \end{pmatrix} \xrightarrow{\jmath^*} W^\perp$$

$$\iota = \begin{pmatrix} \iota_+ \\ \iota_- \end{pmatrix} \quad \jmath = \begin{pmatrix} \cancel{\jmath_+} \\ \cancel{\jmath_-} \end{pmatrix} \begin{pmatrix} \jmath_+ \\ \jmath_- \end{pmatrix}$$

$$\iota^* \jmath = 0$$

$$\jmath^* \iota = 0$$

$$\iota^*_\iota = (\iota_+^* \ \iota_-^*) \begin{pmatrix} \iota_+ \\ \iota_- \end{pmatrix} = \iota_+^* \iota_+ + \iota_-^* \iota_- = 1_W$$

$$\jmath^* \jmath = (\jmath_+^* \ \jmath_-^*) \begin{pmatrix} \jmath_+ \\ \jmath_- \end{pmatrix} = \jmath_+^* \jmath_+ + \jmath_-^* \jmath_- = 1_{W^\perp}$$

$$\begin{pmatrix} V_+ \\ V_- \end{pmatrix} \xleftrightarrow{\iota_+^* \ j_+} \begin{pmatrix} W \\ W^\perp \end{pmatrix} \xleftrightarrow{\iota_-^* \ j_-^*} \begin{pmatrix} V_+ \\ V_- \end{pmatrix}$$

$SL_2$  stuff

$$W \xleftarrow{\iota} \begin{pmatrix} V_+ \\ V_- \end{pmatrix} \xrightarrow{\jmath} W^\perp$$

Recall the ~~case~~ general

$$\begin{pmatrix} X \\ Y \end{pmatrix} \xleftarrow{\sim} \begin{pmatrix} U \\ V \end{pmatrix}$$

should be the same as ~~a short exact sequence~~

equipped with a contraction.

$$0 \longrightarrow X \longrightarrow Y \longrightarrow Z \longrightarrow 0$$

$$y \leftarrow \begin{pmatrix} X \\ Y \end{pmatrix} \leftarrow u$$

TFAE: when

$$\begin{pmatrix} u \\ v \end{pmatrix} \sim \begin{pmatrix} x \\ y \end{pmatrix}$$

$$\begin{pmatrix} a \\ b \end{pmatrix} \begin{pmatrix} h \\ i \end{pmatrix} \begin{pmatrix} f \\ g \end{pmatrix}$$

iff  
equiv. of length 1  
complexes

Review this quickly. Suppose given a short exact sequence

$$V \xleftarrow{\begin{pmatrix} c & d \\ f & g \end{pmatrix}} \begin{pmatrix} X \\ Y \end{pmatrix} \xrightarrow{\begin{pmatrix} a & b \\ h & i \end{pmatrix}} U$$

together with

a splitting:

$$ca + db = 0^v$$

$$ha + ib = 1_u^v$$

$$hf + ig = 0^v$$

$$cf + dg = 1_v^v$$

$$1_x = fc + ah^* \quad 0 = gc + bh^*$$

$$1_y = gd + bi^* \quad 0 = fd + ci^*$$

$$\begin{pmatrix} X \\ Y \end{pmatrix} \xleftarrow{\begin{pmatrix} a & b \\ f & g \end{pmatrix}} \begin{pmatrix} u \\ v \end{pmatrix} \xleftarrow{\begin{pmatrix} h & i \\ c & d \end{pmatrix}} \begin{pmatrix} X \\ Y \end{pmatrix} \xleftarrow{\begin{pmatrix} a & b \\ f & g \end{pmatrix}} \begin{pmatrix} u \\ v \end{pmatrix}$$

$$\begin{pmatrix} 1_u & 0 \\ 0 & 1_v \end{pmatrix}$$

$$\begin{matrix} V & \xleftarrow{(c \ d)} & \begin{pmatrix} X \\ Y \end{pmatrix} & \xleftarrow{(a \ b)} & U \\ & \xrightarrow{(f \ g)} & & \xrightarrow{(h \ i)} & \end{matrix}$$

$$ca+db=0 \checkmark$$

$$hf+ig=0 \checkmark$$

$$V = (c \ d) \begin{pmatrix} f \\ g \end{pmatrix} = cf + dg \quad \boxed{ } \quad \boxed{l_u = (h \ i) \begin{pmatrix} a \\ b \end{pmatrix} = ha + ib}$$

$$\begin{pmatrix} l_x \\ l_y \end{pmatrix} = \begin{pmatrix} f \\ g \end{pmatrix} (c \ d) + \begin{pmatrix} a \\ b \end{pmatrix} (h \ i) \quad \cancel{\boxed{}}$$

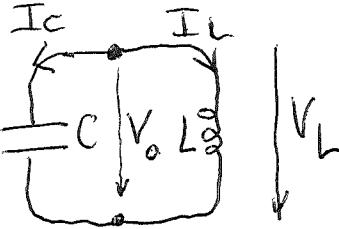
$$\begin{pmatrix} l_x & 0 \\ 0 & l_y \end{pmatrix} = \begin{pmatrix} fc + ah & fd + ai \\ gc + bh & gd + bi \end{pmatrix}$$

$$\begin{pmatrix} X \\ Y \end{pmatrix} \xleftarrow{(a \ b \ f \ g)} \begin{pmatrix} U \\ V \end{pmatrix} \xleftarrow{(h \ i \ c \ d)} \begin{pmatrix} X \\ Y \end{pmatrix}$$

$$\begin{pmatrix} U \\ V \end{pmatrix} \xleftarrow{(h \ i \ c \ d)} \begin{pmatrix} X \\ Y \end{pmatrix} \xleftarrow{(a \ b \ f \ g)} \begin{pmatrix} U \\ V \end{pmatrix}$$

It seems like 8 equations in 8 unknowns

2 do simple oscillation



Kirchhoff

$$V_C = V_L$$

$$I_C + I_L = 0$$

Ohm

$$CV_C = I_C, LI_L = V_L$$

$$\ddot{V}_C = \frac{1}{C} \dot{I}_C = -\frac{1}{C} \dot{I}_L = -\frac{1}{CL} \dot{V}_L = -\frac{1}{CL} V_C, (\frac{\partial^2}{t^2} + \frac{1}{CL}) V_C = 0$$

$$\frac{V_o}{C} \xrightarrow{O} \begin{pmatrix} V_o \\ V_L \end{pmatrix} = \begin{pmatrix} V_o \\ V_o \end{pmatrix}, \begin{pmatrix} V_C \\ V_L \end{pmatrix} \xrightarrow{H} V_\lambda = V_C - V_L$$

$$\begin{pmatrix} C_s & 0 \\ 0 & \frac{1}{L_s} \end{pmatrix} \xrightarrow{G} \begin{pmatrix} \frac{1}{C_s} & 0 \\ 0 & L_s \end{pmatrix}$$

$$\bar{C}_o \longleftrightarrow G_1 \longleftrightarrow H_1$$

$$I_o = I_C + I_L \longleftrightarrow \begin{pmatrix} I_C \\ I_L \end{pmatrix}, \begin{cases} I_C = I_\lambda \\ I_L = -I_\lambda \end{cases} \longleftrightarrow I_\lambda$$

$$V_o \xrightarrow{O} \begin{pmatrix} V_o \\ V_o \end{pmatrix} \xrightarrow{H} \left( \begin{pmatrix} \frac{1}{C_s} & I_\lambda \\ -L_s & I_\lambda \end{pmatrix} \right) \xrightarrow{G} \left( \frac{1}{C_s} + L_s \right) I_\lambda = V_\lambda$$

$$\begin{pmatrix} I_o \\ (C_s + \frac{1}{L_s})V_o \end{pmatrix} \xleftarrow{G} \begin{pmatrix} C_s V_o \\ \frac{1}{L_s} V_o \end{pmatrix} \xleftarrow{H} \begin{pmatrix} I_\lambda \\ -I_\lambda \end{pmatrix} \xleftarrow{I} I_\lambda$$

Thus  $I_\lambda \mapsto I_\lambda V_\lambda = I_\lambda \left( \frac{1}{C_s} + L_s \right) I_\lambda$  is the quadratic form on  $H_1$ . Upon applying the isom.  $I_\lambda \mapsto \left( \frac{1}{C_s} + L_s \right) I_\lambda = V_\lambda$  one gets the quadratic form  $V_\lambda \mapsto V_\lambda \frac{1}{\frac{1}{C_s} + L_s} V_\lambda$  on  $H'$

$\beta$   $V$  2dm Euclidean

equipped with orthogonal involution  $\varepsilon, F$ . Write  $V = \begin{pmatrix} V_+ \\ V_- \end{pmatrix}$   $\varepsilon = \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix}$

Then  $F\varepsilon = g$  is a rotation

$$F\varepsilon = \begin{pmatrix} \cos\theta & -\sin\theta \\ \sin\theta & \cos\theta \end{pmatrix}$$

through a definite angle  $\theta$ .

$$F = \begin{pmatrix} \cos\theta & +\sin\theta \\ \sin\theta & -\cos\theta \end{pmatrix}$$

$\overbrace{V}$  f.d. Euclidean with  $F, \varepsilon$

$$\varepsilon = \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix} \text{ on } V = \begin{pmatrix} V_+ \\ V_- \end{pmatrix}$$

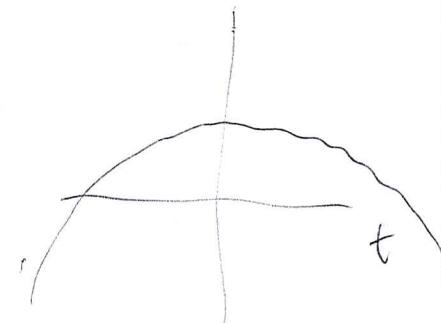
Suppose  $\frac{F\varepsilon + \varepsilon F}{2} = (\cos\theta) \mathbf{Id}$

So you look at

$$\begin{bmatrix} \frac{g+g^{-1}}{2} & \frac{g-g^{-1}}{2} \\ \frac{g-g^{-1}}{2} & \frac{g+g^{-1}}{2} \end{bmatrix}$$

Legendre transform.

$$\int e^{-st + F(t)} \hat{F}(s)$$



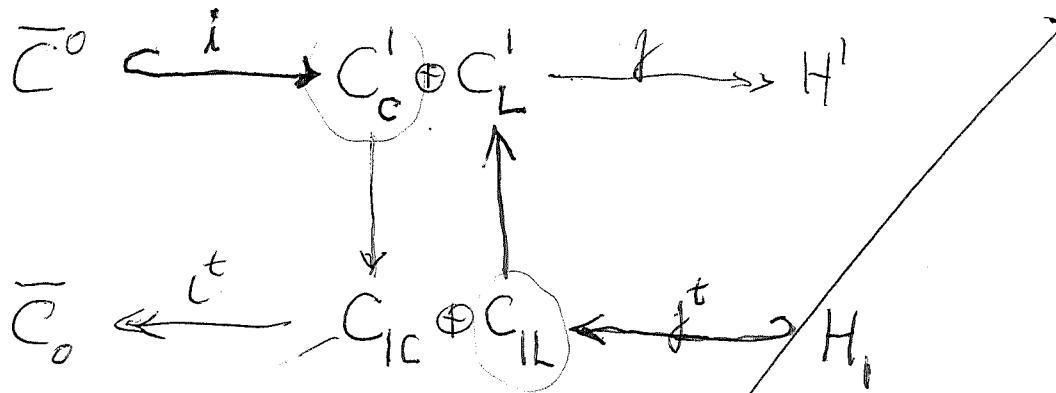
critical point occurs at  $\frac{d}{dt} (-st + F(t)) = -s + F'(t) = 0$

$F'(t) = s$  Use  $F'(t) = s$  to make  $t$  a function of  $s$ . Then treat  $\hat{F}(s) = -st + F(t)$  as a function of  $s$

$$\frac{d}{ds} \hat{F}(s) = -t - s \cancel{\frac{dt}{ds}} + F'(t) \cancel{\frac{dt}{ds}}$$

8 Still where do you get dynamics -

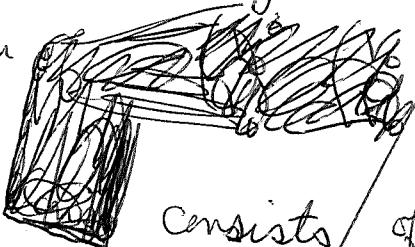
Consider general case



~~What does~~ Write down equations. variables  $V_C$   $I_C$   
 $V_L$   $I_L$

~~defining~~ The point is transversality of the Kirchhoff space and  $\Gamma_s$ . You know that  $K \cap \Gamma_s \neq \emptyset$  means  $\exists$  ~~space~~ ~~subset~~ normal mode of frequency  $s$ . Let's work it out for the simple oscillator. Kirchhoff space is  $(V^*) \oplus I_2$

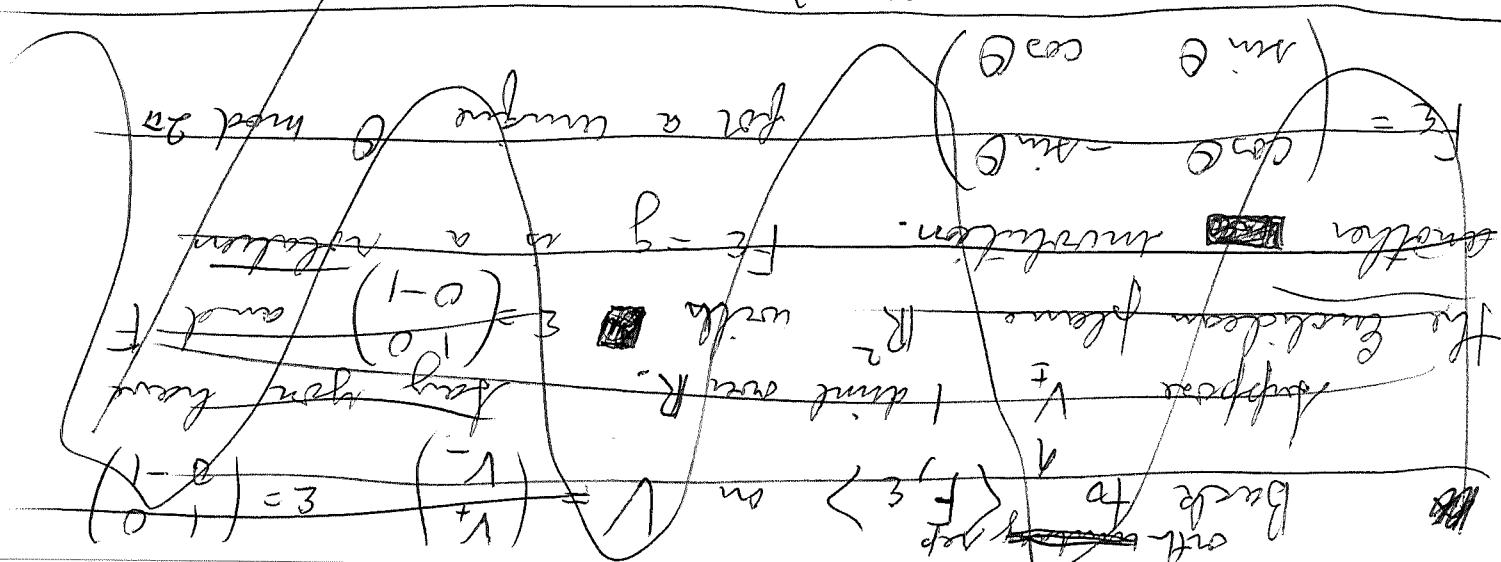
except that you should express it as the direct sum



$$\left\{ \begin{array}{l} V_C \\ V_L \end{array} \right\} \mid V_C = V_L \} \oplus \left\{ \begin{array}{l} I_C \\ I_L \end{array} \right\} \mid I_C + I_L = 0 \}$$

consists of

$$\left\{ \begin{array}{l} V_C \\ C_s V_C \end{array} \right\} \oplus \left\{ \begin{array}{l} V_L \\ \frac{1}{L_s} V_L \end{array} \right\}$$



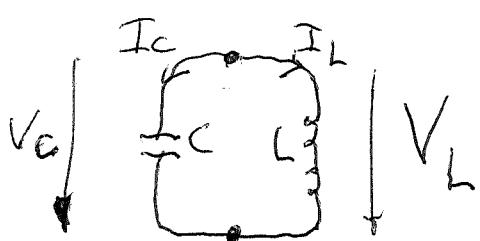
8 Put these ~~equations together~~ Conditions together

$$V_C = V_L \quad I_C + I_L = 0 \quad C_s V_C = I_C, \quad L_s I_L = V_L$$

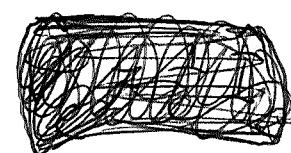
$$C_s V_C = C_s V_L = C_s L_s I_L = -C_s L s^2 I_C = -C_s L s^2 C_s V_C$$

$$(1 + C_s L s^2) V_C = 0$$

Let's next do the IVP.



$$\frac{d}{dt} V_C = \frac{1}{C} I_C \quad \frac{d}{dt} I_L = L V_L$$



$$s \hat{V}_C - V_C(0) = \frac{1}{C} \hat{I}_C = -\frac{1}{C} \hat{I}_L$$

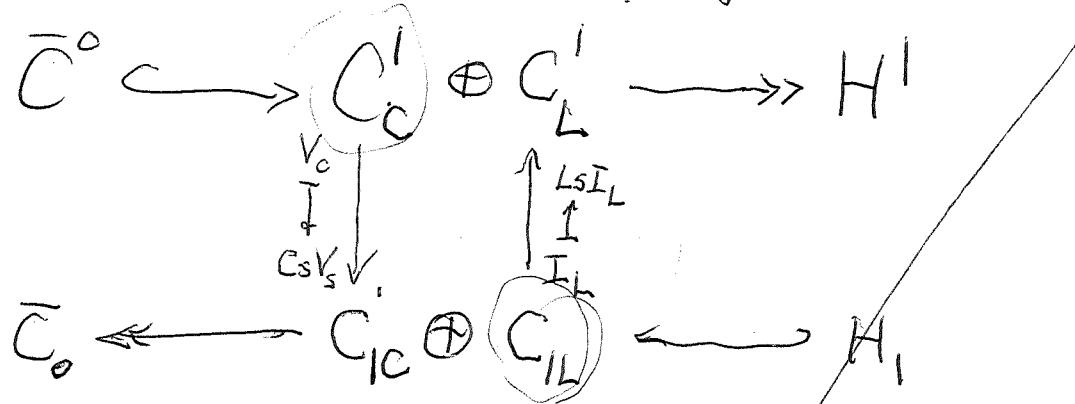
$$s \hat{I}_L - I_L(0) = \frac{1}{L} \hat{V}_L = \frac{1}{L} \hat{V}_C$$

$$\begin{pmatrix} s & \frac{1}{C} \\ -\frac{1}{L} & s \end{pmatrix} \begin{pmatrix} \hat{V}_C \\ \hat{I}_L \end{pmatrix} = \begin{pmatrix} V_C(0) \\ I_L(0) \end{pmatrix}$$

$$\begin{pmatrix} \hat{V}_C \\ \hat{I}_L \end{pmatrix} = \frac{1}{s^2 + \frac{1}{LC}} \begin{pmatrix} s & -\frac{1}{C} \\ \frac{1}{L} & s \end{pmatrix} \begin{pmatrix} V_C(0) \\ I_L(0) \end{pmatrix}$$

done Oct 8, 2002.

E Let's next look at the general case.



~~$\dot{V}_C(t) = \frac{1}{C} \bar{I}_C(t)$~~ 
 $\dot{I}_L(t) = \frac{1}{L} V_L(t)$

$$s \hat{V}_C - V_C(0) = \frac{1}{C} \hat{I}_C \quad s \hat{I}_L - I_L(0) = \frac{1}{L} \hat{V}_L$$

So you ~~get~~ get the equations. ↑ there are e of these, which have to be combined with Kirchhoff's constraints for a total of 2e equations.

Main claim is the general case splits orthogonally into simple oscillator types (4 diml symph phase space) with 2 diml constraints.

Idea now is to take a Grassmannian situation

$$W \hookrightarrow \begin{pmatrix} V_+ \\ V_- \end{pmatrix} \longrightarrow W^\perp$$

$$\begin{pmatrix} 5 & 0 \\ 0 & 5^{-1} \end{pmatrix} \downarrow$$

$$W^* \leftarrow \begin{pmatrix} V_+^* \\ V_-^* \end{pmatrix} \leftarrow (W^\perp)^*$$

5

Recall yesterday's insight

$$\begin{array}{ccccc} \bar{C}^0 & \hookrightarrow & C^1 & \longrightarrow & H^1 \\ & & \uparrow \Gamma_s & & \\ & & \bar{C}_0 & \longleftarrow & C_1 \longrightarrow H_1 \end{array}$$

~~The standard case~~ The regular case is where the Kirchhoff space and  $\Gamma_s$  are transversal. When this fails i.e.  $K \cap \Gamma_s \neq \emptyset$ , then you have normal modes (free motion).

In the regular case  $K \oplus \Gamma_s \cong C^1 \oplus C_1$

Other point is that the L.T. solves the IVP.

~~Then~~ The L.T. of  $\dot{X}(t)$  is  $s\hat{X}(s) - X(0)$ , so the initial values of the dominant variables give rise to and inhomogeneous terms.

~~Now~~ Thing to do next is to decompose the situation into simple oscillators. To go back over the Grassmannian situation orth repn of  $\langle F, \varepsilon \rangle$  on Euclidean space  $V$ .

$$V = \begin{pmatrix} V^+ \\ V^- \end{pmatrix} \quad \varepsilon = \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix}$$

You want to ~~go over~~ go over things you have tried

$$W \xrightarrow{\begin{pmatrix} X \\ Y \end{pmatrix}} \begin{pmatrix} V^+ \\ V^- \end{pmatrix}$$

$$\gamma \quad W = \begin{pmatrix} x \\ y \end{pmatrix} R \quad W^\perp = \begin{pmatrix} -y \\ x \end{pmatrix} R$$

$$F \begin{pmatrix} x & -y \\ y & x \end{pmatrix} = \begin{pmatrix} x & -y \\ y & x \end{pmatrix} \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix}$$

$$F \begin{pmatrix} 1 & -T^* \\ T & 1 \end{pmatrix} = \begin{pmatrix} 1 & -T^* \\ T & 1 \end{pmatrix} \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix}$$

$$F(1+x) = (1+x)\varepsilon = \varepsilon(1+x)$$

$$\frac{1+x}{1-x} = F\varepsilon$$

$$F = \begin{pmatrix} x & -y \\ y & x \end{pmatrix} \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix} \frac{1}{x^2+y^2} \begin{pmatrix} x & y \\ -y & x \end{pmatrix}$$

Setup 2 dim case

~~$\text{W} \hookrightarrow \begin{pmatrix} V_+ \\ V_- \end{pmatrix} \rightarrow W^\perp$~~

~~$R \begin{pmatrix} (x) \\ (y) \end{pmatrix} \xrightarrow{\begin{pmatrix} R \\ R \end{pmatrix}} \begin{pmatrix} R \\ R \end{pmatrix} \xrightarrow{(-y+x)} R$~~

transversality means

$$(x \ y) \begin{pmatrix} s & 0 \\ 0 & s^{-1} \end{pmatrix} \begin{pmatrix} x \\ y \end{pmatrix} \neq 0$$

but you want to  
~~not~~ find the splitting

$$R \xleftarrow{(x \ y)} \begin{pmatrix} R \\ R \end{pmatrix} \xleftarrow{(-y \ x)} R$$

$$s\hat{V}_C - V_C(0) = 1$$

$$V_o \rightarrow \begin{pmatrix} xV_o \\ yV_o \end{pmatrix}$$

Maybe you want  
normalized variables

Call them  $V_c, V_L$

$$x\dot{V}_c = xI_c = -yI_L$$

$$x\dot{I}_L = xV_L = yV_c$$

$$\dot{V}_c = -\left(\frac{y}{x}\right)I_L$$

$$\dot{I}_L = \left(\frac{y}{x}\right)V_c$$

~~equations~~ are

$$\dot{V}_c = I_c$$

$$\dot{I}_L = V_L$$

$$xI_c + yI_L = 0$$

$$-yV_c + xV_L = 0$$

~~$\dot{V}_c = -\frac{y}{x}\dot{I}_L$~~ 
 ~~$\dot{I}_L = \frac{y}{x}V_c$~~

here  $\frac{y}{x} = \frac{\sin}{\cos} = \tan$

$$W = \begin{pmatrix} 1 \\ T \end{pmatrix} V_C \begin{pmatrix} V_+ \\ V_- \end{pmatrix} \xrightarrow{T \rightarrow V_-}$$

$$\dot{V}_c = I_c$$

$$\dot{I}_L = V_L$$

$$-TV_C + V_L = 0$$

$$I_C + T^* I_L = 0$$

$$W \xleftarrow{(1-T^*)} \begin{pmatrix} V_+ \\ V_- \end{pmatrix} \xleftarrow{(-T^*)} V_-$$

$$s\hat{V}_C - V_C(0) = \hat{I}_C = -T^*\hat{I}_L$$

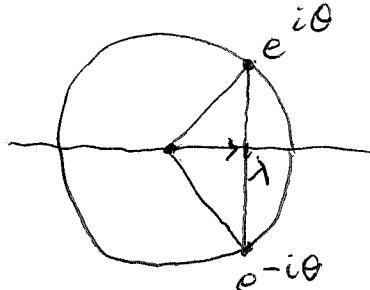
$$s\hat{I}_L - I_L(0) = \hat{V}_L = T\hat{V}_C$$

$$\begin{pmatrix} s & T^* \\ -T & s \end{pmatrix} \begin{pmatrix} \hat{V}_C \\ \hat{I}_L \end{pmatrix} = \begin{pmatrix} V_C(0) \\ I_L(0) \end{pmatrix}$$

Next you want the splitting of ~~an~~ an LC phase space into simple oscillators.

Let  $V = \begin{pmatrix} V_+ \\ V_- \end{pmatrix}$  be a polarized Hilb space,  $\Sigma = \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix}$ . Let  $F$  be a unitary involution with  $\pm 1$  eigenspaces  $W$  and  $W^\perp$  resp. Then  $g = F\Sigma$  is unitary on  $V$ . Also  $\frac{g+g^{-1}}{2}$  is selfadjoint and it commutes with both  $F, \Sigma$ , so that the ~~the~~ representation of  $F, \Sigma$  on  $V$  decomposes orthogonally  $V = \bigoplus_{\lambda} V_\lambda$  where  $V_\lambda$  is the  $\lambda$  eigenspace of  $\frac{g+g^{-1}}{2}$  and  $-1 < \lambda < 1$ . (For the moment we ignore the  $\pm 1$  eigenspaces of  $g$ .)

Suppose  $V = V_\lambda$  i.e.  $g$  satisfies  $\frac{g+g^{-1}}{2} = \lambda$  (scalar op.). Roots of  $\frac{z+z^{-1}}{2} = \lambda$ , i.e.  $z^2 + 2\lambda z + 1$ , are  $\lambda \pm \sqrt{\lambda^2 - 1}$ , i.e.  $\cos \theta \pm i \sin \theta$ , where  $\cos \theta = \lambda$ .



Since  $(g - e^{i\theta})(g - e^{-i\theta}) = g^2 - 2\lambda g + 1 = 0$

$$(g - e^{-i\theta}) + (-g + e^{i\theta}) = 2i \sin \theta.$$

So  $\frac{g - e^{-i\theta}}{2i \sin \theta}, \frac{-g + e^{i\theta}}{2i \sin \theta}$  are annihilating

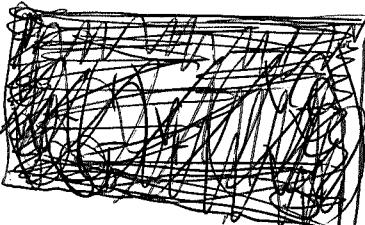
idempotents with sum 1 on  $V_\lambda$ .

K Consider  $V$  (Hilb op.),  $F, \varepsilon$  invs

assume  $\frac{Fe + \varepsilon F}{2} = \cos \theta \quad 0 < \theta < \pi$ . Put  $g = Fe$ .

~~then~~  $(g - e^{i\theta})(g - e^{-i\theta}) = 0$ , you get  
ans. idemp.  $e_{-\theta} = \frac{-g + e^{i\theta}}{2i \sin \theta}$   $e_{+\theta} = \frac{g - e^{-i\theta}}{2i \sin \theta}$

Then  $V = \begin{pmatrix} V_0 \\ V_{-\theta} \end{pmatrix} \quad \varepsilon e_{-\theta} \varepsilon = \frac{g^{-1} e^{-i\theta}}{2i \sin \theta} ?$



$$W = \begin{pmatrix} 1 \\ T \end{pmatrix} V_+ \subset \begin{pmatrix} V_+ \\ V_- \end{pmatrix}$$

$$W^\perp = \begin{pmatrix} -T^* \\ 1 \end{pmatrix} V_- \subset \begin{pmatrix} V_+ \\ V_- \end{pmatrix}$$

$$F \begin{pmatrix} 1 & -T^* \\ T & 1 \end{pmatrix} = \begin{pmatrix} 1 & T^* \\ T & 1 \end{pmatrix} \varepsilon = \varepsilon(1-X)$$

②  $\frac{1+X}{1-X} = Fe = g. \quad g = -1 \Leftrightarrow X = \infty.$

This is the Cayley transform theory. ~~It works over R~~

It works over  $\mathbb{R}$ , where  $g$  is orthogonal

Important idea is how to quantize a harmonic oscillator. On phase space you have a symplectic form  $\omega$ , ~~and a metric~~. You also have a symmetric bilinear form  $H$ , the ~~Hamiltonian~~  $X$ . These combine to give you a skewsymmetric form  $X$ .

Quantization uses polar decomp:  $X = \sqrt{|X|} J$   
 $J^2 = -1$ .  $|X|$  gives the energy levels  $J$  the guarantee ~~of~~ kinematics.

$\lambda$  Program now is make explicit this stuff

Return to  $V, F, \varepsilon$ . Your aim is decompose into isotypical components and to put these in a canonical form. You want to make any choices explicit.  $\# g = F\varepsilon$  ~~uses~~ the order of  $F, \varepsilon$  also  $\cos \theta$  ~~also~~ uses  $\theta \in (0, \pi)$ .

Hilbert case. Assume  $\frac{g+g^{-1}}{2} = \cos \theta$  on  $V$ . Then  $V = V_\theta \oplus V_{-\theta}$  where  $V_\theta = \{ \xi \mid g\xi = e^{i\theta}\xi \}$ . ~~also~~ ~~rank~~ ~~isom with inverse~~  $\varepsilon$  restricted to  $V_\theta$  is an isom  $\varepsilon: V_\theta \xrightarrow{\sim} V_{-\theta}$  with inverse given by  $\varepsilon^{-1}$ .

$$g\{\xi\} = e^{i\theta}\{\xi\}$$

$$g^{-1}\{\varepsilon\} = \bar{e}^{-i\theta}\{\varepsilon\}$$

~~rank~~

What's the relation between  $V_\theta, V_{-\theta}, V_+, V_-, W, W^\perp$

$$\begin{pmatrix} V_\theta \\ V_{-\theta} \end{pmatrix} \not\sim \begin{pmatrix} V_+ \\ V_- \end{pmatrix} \simeq \begin{pmatrix} W \\ W^\perp \end{pmatrix} \quad \text{~~rank~~}$$

So you have:  $g \in F$  ~~giving~~ yielding these.  $(g - e^{i\theta})(g - \bar{e}^{-i\theta}) = g^2 - (2\cos \theta)g + 1 = 0$   
 $\Leftrightarrow g + g^{-1} - 2\cos \theta = 0$ .

$$(g - e^{-i\theta}) + (g + e^{+i\theta}) = +2i\sin \theta$$

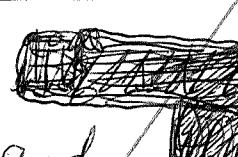
$$\frac{g - e^{-i\theta}}{2i\sin \theta} + \frac{e^{i\theta} - g}{2i\sin \theta} = 1$$

$\mu$

$$\pi_\theta = \frac{g - e^{-i\theta}}{2i \sin \theta}$$

$$\pi_{-\theta} = \frac{e^{i\theta} - g}{2i \sin \theta}$$

Review: Aim? Given  $V$  repn of  $F, \varepsilon$   
 sat  $\frac{F_\varepsilon + \varepsilon F}{2} = \cos \theta$   $\theta \in (0, \pi)$ . You  
 want a canonical picture  for  $V$ . What  
 might this be?  $V$  splits into irreducibles  
 which are all isomorphic ~~also~~, and the endom.  
 ring is  $\mathbb{C}$ .

IDEA. similarity between  roots of  
 an irreducible equation and  random  
 phases

Back to irreducible unitary repns of the inf  
 dihedral group. Should be induced from a  
 character on the infinite cyclic group by Mackey  
 theory. ~~Character~~ Set

$$g \mapsto \begin{pmatrix} e^{i\theta} & 0 \\ 0 & e^{-i\theta} \end{pmatrix} \quad \varepsilon \mapsto \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \text{ on } \mathbb{C}^2.$$

image of  $\langle F, \varepsilon \rangle$  in  $M_2(\mathbb{C})$  generates. Endo  
 ring of the rep is  $\mathbb{C}$

Next consider orthogonal repns of  $F, \varepsilon$  on a  
 Euclidean space. ~~orthogonal~~

$$g = \begin{pmatrix} \cos \theta & -\sin \theta \\ \sin \theta & \cos \theta \end{pmatrix} \quad \varepsilon = \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix} \text{ on } \mathbb{R}^2$$

Again  $\langle F, \varepsilon \rangle$  generate  $M_2(\mathbb{R})$   $\therefore$  irred

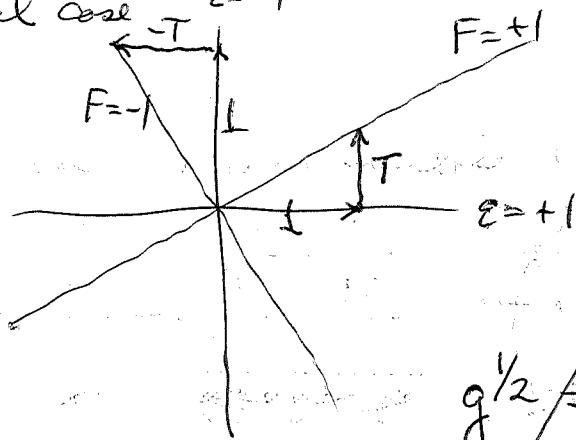
So for any repn of  $F_\varepsilon$  with canonical  
 $\frac{1}{2}(F_\varepsilon + \varepsilon F) = \cos \theta$  one has an isom

$$V \cong P \otimes \text{Hom}_{F_\varepsilon}(P, V)$$

unique up to a scalar factor of  $|1|=1$ .

 You now need to link these ideas from repn theory to other things.

real case  $\varepsilon = -1$



$$F \begin{pmatrix} 1 & -T^* \\ T & 1 \end{pmatrix} = (1+X)\varepsilon = \varepsilon(1-X)$$

$$\frac{1+X}{1-X} = F\varepsilon = g.$$

$$g^{1/2} = \frac{1+X}{\sqrt{1-X^2}} = \begin{pmatrix} 1 & -T^* \\ T & 1 \end{pmatrix} \frac{1}{\sqrt{(1-T^*)^2}} = \begin{pmatrix} \frac{1}{\sqrt{(1-T^*)^2}} & -T^* \frac{1}{\sqrt{(1+TT^*)^2}} \\ T \frac{1}{\sqrt{(1-T^*)^2}} & \frac{1}{\sqrt{(1+TT^*)^2}} \end{pmatrix}$$

Consider real case  $V, F, \varepsilon$   $W = \begin{pmatrix} 1 \\ T \end{pmatrix} v_+$ ,  $W^\perp = \begin{pmatrix} -T^* \\ 1 \end{pmatrix} v_-$

$$F \begin{pmatrix} 1 & -T^* \\ T & 1 \end{pmatrix} = \begin{pmatrix} 1 & -T^* \\ T & 1 \end{pmatrix} \varepsilon \quad F(1+X) = (1+X)\varepsilon = \varepsilon(1-X)$$

$$\frac{1+X}{1-X} = F\varepsilon = g. \quad \text{Assume } \frac{g+g^{-1}}{2} = \cos \theta \quad \theta \in (0, \pi)$$

$$\cos \theta = \left( \frac{1+X}{1-X} + \frac{1-X}{1+X} \right) \frac{1}{2} = \frac{1+X^2}{1-X^2} = \begin{pmatrix} 1 & 1 \\ -1 & 1 \end{pmatrix} X^2 \Rightarrow X^2 = \frac{\cos \theta - 1}{\cos \theta + 1}$$

$$-X^2 = \frac{1-\cos \theta}{1+\cos \theta} = \frac{\sin^2(\theta/2)}{\cos^2(\theta/2)} \quad \text{but } -X^2 = -\begin{pmatrix} 0 & -T^* \\ T & 0 \end{pmatrix} \begin{pmatrix} 0 & -T^* \\ T & 0 \end{pmatrix}$$

$$-X^2 = \begin{pmatrix} T^* & 0 \\ 0 & TT^* \end{pmatrix} \quad \therefore -X^2 = \left( \frac{\sin \theta/2}{\cos \theta/2} \right)^2$$

something happens here with the phase of  $T$ .  $T^*T$  and  $T^*T = (\tan \frac{\theta}{2})^2$  so the char values of  $T$  are all  $= |\tan \frac{\theta}{2}|$  which ranges over  $(0, \infty)$  for  $\theta \in (0, \pi)$ . The eigenvalues of  $(\begin{smallmatrix} 0 & -T^* \\ T & 0 \end{smallmatrix})$  are purely imaginary.

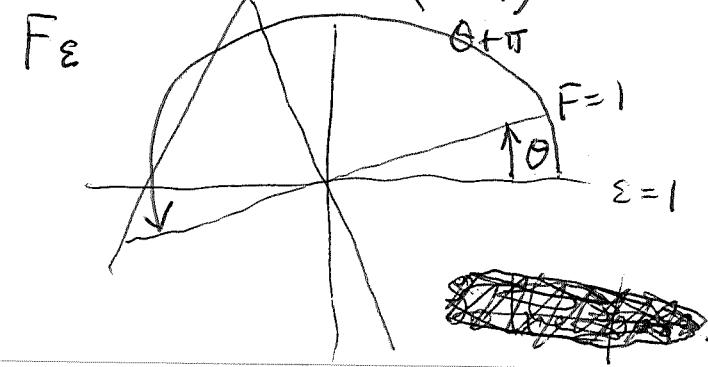
You have seen that the LC circuit resolvent is  $\begin{pmatrix} s & T^* \\ -T & s \end{pmatrix}^{-1}$  which leads to singularities of  $\frac{s}{s^2 + \omega^2}$ .

$$g^{1/2} = \frac{1+x}{\sqrt{1-x^2}} = \begin{pmatrix} 1 & -T^* \\ -T & 1 \end{pmatrix} \begin{pmatrix} \frac{1}{\sqrt{1-T^*T}} & 0 \\ 0 & \frac{1}{\sqrt{1-T^*T}} \end{pmatrix}$$

Something here should contain the phase  $T$  of  $X = \begin{pmatrix} 0 & -T^* \\ T & 0 \end{pmatrix}$ , namely, replace  $X$  by  $tX$  and let  $t \rightarrow \infty$ .

Aim 12 Oct is to sort out ~~erroneous~~  $F, \varepsilon$  reps.

Idea: There is this angle  $\theta$  whose double is intrinsic? Let  $V$  be an ~~wired~~ <sup>orthogonal</sup> repn of  $F, \varepsilon$  in  $\mathbb{R}^2$  such that  $\varepsilon = \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix}$ . What are the possibilities for  $F$ ?



$$\text{Put } g^{1/2} = \begin{pmatrix} \cos \theta & -\sin \theta \\ \sin \theta & \cos \theta \end{pmatrix}$$

$$g^{1/2} \varepsilon g^{-1/2} = g\varepsilon = \begin{pmatrix} \cos 2\theta & -\sin 2\theta \\ \sin 2\theta & \cos 2\theta \end{pmatrix}$$

$$\text{Possible } F = g^{1/2} \varepsilon g^{-1/2}$$

Vague Idea: There seems to be a loss of information involved in passing from an irreducible representation of  $F, \varepsilon$  on  $\mathbb{R}^2$  to the corresponding harmonic oscillator. The oscillator retains only positive type information, e.g. characteristic values. Now quantization of the oscillator involves lifting to a double covering, that is, undoing the previous step in some way.

Start with  $V = \mathbb{R}^2$   $\varepsilon = \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix}$  and  $F$  another orthogonal involution.

Set  $g^{1/2} = \begin{pmatrix} \cos \theta & -\sin \theta \\ \sin \theta & \cos \theta \end{pmatrix}$

Then  $g^{1/2} \varepsilon g^{-1/2} =$

$$\begin{pmatrix} \cos \theta & \sin \theta \\ \sin \theta & -\cos \theta \end{pmatrix} \begin{pmatrix} \cos \theta & \sin \theta \\ -\sin \theta & \cos \theta \end{pmatrix}$$

$$= \begin{pmatrix} \cos(2\theta) & \sin(2\theta) \\ \sin(2\theta) & -\cos(2\theta) \end{pmatrix}$$

Simpler would have been  $g^{1/2} \varepsilon g^{-1/2} = g\varepsilon$

Let's go over this again. You are looking at wired orth rep of the infinite dihedral group  $\langle F, \varepsilon \rangle$  on  $\mathbb{R}^2$  such that  $\varepsilon = \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix}$ . Then  $g = F\varepsilon$  is an orthogonal transformation of  $\det = +1$  (orientation-preserving)

so it's a rotation  $g = \begin{pmatrix} \cos 2\theta & -\sin 2\theta \\ \sin 2\theta & \cos 2\theta \end{pmatrix}$ , get two values

$$\text{for } g^{\frac{1}{2}} = \begin{pmatrix} \cos \theta & -\sin \theta \\ \sin \theta & \cos \theta \end{pmatrix} ?$$

Confused again. Given orth rep of  $F, \varepsilon$  on  $\mathbb{R}^2$   
s.t.  $\varepsilon = \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix}$ . Then  $g = F\varepsilon$  is a rotation, i.e.  
 $g \in SO(2, \mathbb{R})$ . There are 2 square roots of  $g$   
namely  $g^{\frac{1}{2}} = \begin{pmatrix} \cos \theta & -\sin \theta \\ \sin \theta & \cos \theta \end{pmatrix}, \begin{pmatrix} \cos \theta & +\sin \theta \\ -\sin \theta & \cos \theta \end{pmatrix} = g^{-\frac{1}{2}}$ ?

Given orth rep of  $F, \varepsilon$  on  $\mathbb{R}^2$   $\varepsilon = \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix}$

$$W: F = 1$$

$$V_+: \varepsilon = 1$$

$$g^{\frac{1}{2}} \varepsilon g^{-\frac{1}{2}} = g \varepsilon = F$$

$$g^{\frac{1}{2}} = \begin{pmatrix} \cos \theta & -\sin \theta \\ \sin \theta & \cos \theta \end{pmatrix}$$

$$F = \begin{pmatrix} \cos 2\theta & +\sin 2\theta \\ \sin 2\theta & -\cos 2\theta \end{pmatrix} \quad -\frac{\pi}{2} < \theta < \frac{\pi}{2}$$

$$\frac{1+F}{2} = \begin{pmatrix} \cos^2 \theta & +\cos \theta \sin \theta \\ \cos \theta \sin \theta & \sin^2 \theta \end{pmatrix} = \begin{pmatrix} \cos \theta \\ \sin \theta \end{pmatrix} \begin{pmatrix} \cos \theta & \sin \theta \end{pmatrix}$$

Also  $W = \begin{pmatrix} 1 \\ t \end{pmatrix} V_+ \quad W^\perp = \begin{pmatrix} -t \\ 1 \end{pmatrix} V_-$

$$F \begin{pmatrix} 1 & -t \\ t & 1 \end{pmatrix} = \begin{pmatrix} 1 & -t \\ t & 1 \end{pmatrix} \varepsilon \Rightarrow \frac{1+x}{1-x} = F\varepsilon = g$$

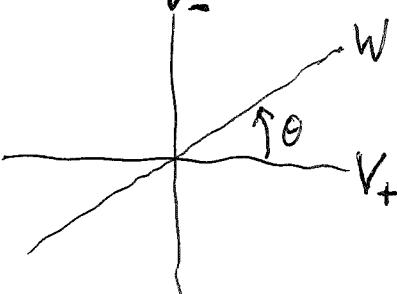
$$1+x \quad 1+x \Rightarrow \frac{1+x}{\sqrt{1-x^2}} = g^{\frac{1}{2}}$$

$$g^{\frac{1}{2}} = \begin{pmatrix} 1 & -t \\ t & 1 \end{pmatrix} \frac{1}{\sqrt{1+t^2}} = \begin{pmatrix} \cos \theta & -\sin \theta \\ \sin \theta & \cos \theta \end{pmatrix}, \quad t = \tan \theta$$

$-\infty < t < \infty$

$-\frac{\pi}{2} < \theta < \frac{\pi}{2}$

Q: Classify irreducible orthogonal reps of  $F, \varepsilon$  on  $\mathbb{R}^2$   
such that  $\varepsilon = \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix}$ . A:



$$F = \begin{pmatrix} \cos \theta & -\sin \theta \\ \sin \theta & \cos \theta \end{pmatrix} \varepsilon \begin{pmatrix} \cos \theta & \sin \theta \\ -\sin \theta & \cos \theta \end{pmatrix}$$

$$F = \begin{pmatrix} \cos 2\theta & \sin 2\theta \\ \sin 2\theta & -\cos 2\theta \end{pmatrix} \quad -\frac{\pi}{2} < \theta < \frac{\pi}{2}$$

Q: Classify irreducible unitary reps of  $F, \varepsilon = \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix}$  on  $\mathbb{C}^2$ . Let  $W = \begin{pmatrix} 1 \\ z \end{pmatrix} V_+$ ,  $W^\perp = \begin{pmatrix} \bar{z} \\ 1 \end{pmatrix} V_-$ ,

$$g^{1/2} = \frac{1+x}{\sqrt{1-x^2}} = \begin{pmatrix} 1 & -\bar{z} \\ z & 1 \end{pmatrix} \frac{1}{\sqrt{1+|z|^2}}$$

$$\text{as } X = \begin{pmatrix} 0 & -\bar{z} \\ z & 0 \end{pmatrix} \Rightarrow 1-X^2 = \begin{pmatrix} 1+|z|^2 & 0 \\ 0 & 1+|z|^2 \end{pmatrix}$$

If  $z = r e^{i\varphi}$ , then

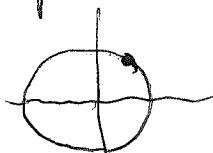
$$g^{1/2} = \begin{pmatrix} \cos \theta & e^{-i\varphi}(-\sin \theta) \\ e^{i\varphi} \sin \theta & \cos \theta \end{pmatrix} \quad \text{where } \frac{1}{\sqrt{1+r^2}} = \cos \theta$$

$$\frac{r}{\sqrt{1+r^2}} = \sin \theta$$

$$\text{so } F = \begin{pmatrix} \cos 2\theta & e^{-i\varphi}(-\sin 2\theta) \\ e^{i\varphi} \sin 2\theta & \cos 2\theta \end{pmatrix}$$

Let  $V$  be a unitary repn of  $F, \varepsilon$  such that  $\frac{1}{2}(g+g^{-1}) = \frac{1}{2}(F\varepsilon + \varepsilon F)$  is the scalar operator  $\lambda I$ , where

$$\lambda = \cos 2\theta \quad -\frac{\pi}{2} < \theta < \frac{\pi}{2}$$



Try to show that  $V = V_{2\theta} \oplus V_{-2\theta}$ , where

$$V_{\pm 2\theta} = \text{Ker}(g - e^{\pm 2i\theta}). \quad \text{You have}$$

$$(g - e^{-2i\theta})(g - e^{2i\theta}) = g^2 - 2\cos(2\theta)g + 1$$

$$= g^2 - 2 \frac{g+g^{-1}}{2}g + 1 = 0$$

$$0 \quad \frac{g - e^{-2i\theta}}{2i \sin 2\theta} + \frac{-g + e^{2i\theta}}{2i \sin 2\theta} = 1$$

OK

so any  ~~$\zeta \in V$~~  splits

$$\underbrace{\frac{g - e^{-2i\theta}}{2i \sin 2\theta} \zeta}_{\text{killed by } g - e^{2i\theta}} + \underbrace{\frac{-g + e^{2i\theta}}{2i \sin 2\theta} \zeta}_{\text{killed by } g - e^{-2i\theta}} = \zeta$$

Also  $\varepsilon \frac{g - e^{-2i\theta}}{2i \sin 2\theta} \varepsilon^{-1} = \frac{g^{-1} - e^{-2i\theta}}{2i \sin 2\theta}$   $\frac{1+\varepsilon}{2}V \quad \frac{1-\varepsilon}{2}V$

so  ~~$V$~~   $V = V_{2\theta} \oplus V_{-2\theta} = \boxed{V_+ \oplus V_-}$

Let  $\zeta \in V_{2\theta} \Rightarrow g\zeta = e^{2i\theta}\zeta \Rightarrow g^{-1}\varepsilon\zeta = e^{2i\theta}\varepsilon\zeta$   
 $\Rightarrow e^{-2i\theta}\varepsilon\zeta = g\varepsilon\zeta \Rightarrow \varepsilon\zeta \in V_{-2\theta}$ . Since

$\zeta \in V_{-2\theta} \Rightarrow \varepsilon\zeta \in V_{2\theta}$ . You now have the rep

$$V = \begin{pmatrix} V_{2\theta} \\ V_{-2\theta} \end{pmatrix} \xrightarrow{g} \begin{pmatrix} e^{2i\theta} & 0 \\ 0 & e^{-2i\theta} \end{pmatrix} \quad \varepsilon \mapsto \begin{pmatrix} 0 & \varepsilon \\ \varepsilon & 0 \end{pmatrix}$$

So perhaps the point is that all of the <sup>sub</sup> spaces  $V_{2\theta}, V_{-2\theta}, V_+, V_-$  are canonically isom.

There is ~~a unique vector space structure~~ ~~on~~ ~~the~~ subspace  $Y$  ~~such that~~ together with an isom  $\mathbb{C}^2 \otimes Y \xrightarrow{\sim} V$  compatible with  $f, \varepsilon$

~~Q: What are the irreducible representations of  $F, \varepsilon$ ?~~

$V$  a unitary rep of  $F, \varepsilon$  such that

$\frac{1}{2}(F\varepsilon + \varepsilon F) = \frac{g+g^{-1}}{2}$  is a scalar operator  $\lambda$ .

then  $\lambda \in [-1, 1]$ , say  $\lambda \in (-1, 1)$

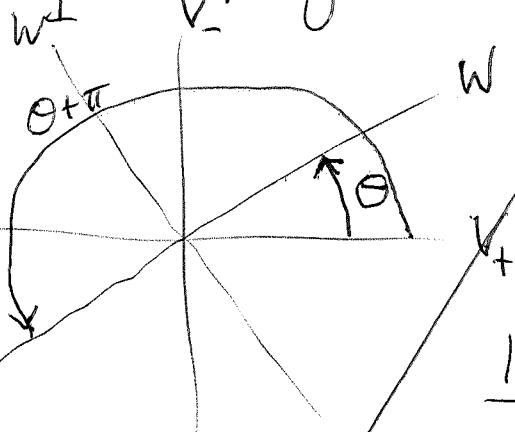
$$\frac{g+g^{-1}}{2} = \lambda \Rightarrow g^2 - 2\lambda g + 1 = 0$$

eigenvalues  $g = \lambda \pm \sqrt{\lambda^2 - 1} = \cos 2\theta \pm i \sin 2\theta$

Then let  $(g - e^{\frac{2i\theta}{2}})(g - e^{-\frac{2i\theta}{2}}) = 0$

$$V = V_{2\theta} \oplus V_{-2\theta} = V_+ \oplus V_-$$

$V$  orth rep of  $F, \varepsilon$  on  $\mathbb{R}^2$   $\varepsilon = \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix}$



$$g^{1/2} = \begin{pmatrix} \cos \theta & -\sin \theta \\ \sin \theta & \cos \theta \end{pmatrix}$$

$$F = g^{1/2} \varepsilon g^{-1/2} = g \varepsilon = \begin{pmatrix} \cos \theta & \sin \theta \\ \sin \theta & -\cos \theta \end{pmatrix}$$

$$\frac{1+F}{2} = \begin{pmatrix} \cos^2 \theta & \sin \theta \cos \theta \\ \sin \theta \cos \theta & \cos^2 \theta \end{pmatrix} = \begin{pmatrix} \cos \theta \\ \sin \theta \end{pmatrix} \begin{pmatrix} \cos \theta & \sin \theta \\ \sin \theta & \cos \theta \end{pmatrix}$$

$$g = \begin{pmatrix} \cos 2\theta & -\sin 2\theta \\ \sin 2\theta & \cos 2\theta \end{pmatrix}$$

$$\frac{g+g^{-1}}{2} = \cos 2\theta$$

Q: What are the irreducible reps.  $\varepsilon = \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix}$ ,

$$F = \begin{pmatrix} \cos 2\theta & \sin 2\theta \\ \sin 2\theta & -\cos 2\theta \end{pmatrix}$$

$$\frac{g+g^{-1}}{2} = \cos 2\theta$$

$$-\frac{\pi}{2} < \theta < \frac{\pi}{2}$$

gives all real lines

$$W = \begin{pmatrix} 1 \\ T \end{pmatrix} V_+ \quad W^\perp = \begin{pmatrix} -T^* \\ 1 \end{pmatrix} V_-$$

$$X = \begin{pmatrix} 0 & -T^* \\ T & 0 \end{pmatrix} \quad -X^2 = \begin{pmatrix} -TT^* & 0 \\ 0 & +TT^* \end{pmatrix}$$

$$F(1+X) = (1+X)\varepsilon = \varepsilon(1-X)$$

$$g = F\varepsilon = \frac{1+X}{1-X} \quad \cancel{\text{cancel}}$$

$$g^{1/2} = \frac{1+X}{\sqrt{1-X^2}} = \begin{pmatrix} 1 & -T^* \\ T & 1 \end{pmatrix} \begin{pmatrix} 1+TT^* & 0 \\ 0 & 1+TT^* \end{pmatrix}^{-1/2}$$

Look at wired unitary reps. in  $\mathbb{C}^2$ .

$$W = \begin{pmatrix} 1 \\ z \end{pmatrix} V_+ \quad W^\perp = \begin{pmatrix} -\bar{z} \\ 1 \end{pmatrix} V_-$$

$$g^{1/2} = \begin{pmatrix} 1 & -\bar{z} \\ z & 1 \end{pmatrix} (1+|z|^2)^{-1/2}$$

$$F = g^{1/2} \varepsilon g^{-1/2} \quad g^{1/4} = \begin{pmatrix} 1 & -\bar{z} \\ z & 1 \end{pmatrix}^2 \frac{1}{1+|z|^2}$$

$$= \begin{pmatrix} 1+|z|^2 & -2\bar{z} \\ 2z & 1-|z|^2 \end{pmatrix} \frac{1}{1+|z|^2} = \begin{pmatrix} \frac{1-|z|^2}{1+|z|^2} & \frac{-2\bar{z}}{1+|z|^2} \\ \frac{2z}{1+|z|^2} & \frac{1-|z|^2}{1+|z|^2} \end{pmatrix}$$

$$F = g\varepsilon = \begin{pmatrix} \frac{1-|z|^2}{1+|z|^2} & \frac{2\bar{z}}{1+|z|^2} \\ \frac{2z}{1+|z|^2} & -\frac{1-|z|^2}{1+|z|^2} \end{pmatrix} = \begin{pmatrix} \cos 2\theta & -e^{i\varphi} \sin 2\theta \\ e^{i\varphi} \sin 2\theta & -\cos 2\theta \end{pmatrix}$$

Maybe what's involved is that the phase  $e^{i\varphi}$  in  $z$  disappears because of something commuting with  $\varepsilon$ ?

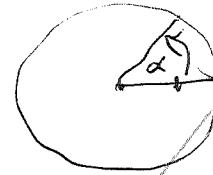
$\varphi$   $F, \varepsilon$  unitary rep on  $V$  given  $\boxed{V = \begin{pmatrix} V_+ \\ V_- \end{pmatrix}} \quad \varepsilon = \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix}$

Assume  $\frac{1}{2}(F\varepsilon + \varepsilon F) = \text{scalar } \cos\alpha$

$$\frac{g+g^{-1}}{2} = \cos\alpha$$

$$\alpha \in (0, \pi)$$

$$-1 < \cos\alpha < 1$$



$$g^2 - (2\cos\alpha)g + 1 = 0 \quad \text{eigen. of } g = \cos\alpha + \sqrt{\cos^2\alpha - 1}$$

$$g + g^{-1} = 2\cos\alpha \quad = \cos\alpha \pm \sqrt{1 - \sin^2\alpha}$$

$$\therefore \text{get decmp. } V = V_+ \oplus V_- \quad V_\alpha = \text{Ker}(g - e^{\frac{i\pi\alpha}{2}})$$

At this point you have

$$\begin{array}{ccc} V_\alpha & \sim & V_+ \\ \oplus & & \oplus \\ V_{-\alpha} & & V_- \end{array}$$

How is  $\overset{\sim}{V}$  related to  $\circledast$  the  $F$  decmp.  $V = \begin{pmatrix} w \\ w^\perp \end{pmatrix}$  ?

Idea: You have 3 decompositions of  $V$  into

$$\begin{array}{ccc} V_\alpha & w & V_+ \\ \oplus & \oplus & \oplus \\ V_{-\alpha} & w^\perp & V_- \end{array}$$

Program. You still want to understand a general unitary repn  $V$  of  $F, \varepsilon$  such that  $\frac{1}{2}(F\varepsilon + \varepsilon F)$  is a scalar.   
 You know that  $\overset{\sim}{V}$  is hermitian of  $\| \cdot \| \leq 1$ , so you have  $\frac{1}{2}(F\varepsilon + \varepsilon F) = \cos\theta$  with  $0 < \theta < \pi$  (ignore  $g = \pm 1$  for the moment). Then you have

$$\frac{1}{2}(g+g^{-1}) - \cos\theta = 0 \quad g^2 - 2g\cos\theta + 1 = 0$$

$$g \text{ has eigenvalues } \cos\theta \pm \sqrt{\cos^2\theta - 1} = \cos\theta \pm i\sin\theta = e^{\pm i\theta}$$

$$(g - e^{i\theta})(g - e^{-i\theta}) = g^2 + g(-2\cos\theta) + 1 = 0$$

X What is the algebra of operators, the image of the group algebra in  $\text{End}(V)$ ? Should be generated by  $F_\varepsilon$  or  $g_\varepsilon$ .

You believe the ~~unique~~ unique wired repn of  $F_\varepsilon$  with  $\frac{1}{2}(g+g^{-1}) = \cos \theta$  is given by ① with  $g = \begin{pmatrix} e^{i\theta} & 0 \\ 0 & e^{-i\theta} \end{pmatrix}$   $\varepsilon = \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix}$

The problem is to find, calculate the  $g$  eigenspaces.  
Perhaps look at  $P^1$

Repeat:  $V$  unitary repn of  $F_\varepsilon$  such that  $\frac{F_\varepsilon + \varepsilon F_\varepsilon^{-1}}{2} = \cos \theta$   
Then you have a splitting  $V = V_g \oplus V_{g^{-1}}$   $g = e^{i\theta}$   
such that  $g = \begin{pmatrix} 1 & 0 \\ 0 & g^{-1} \end{pmatrix}$  on  $(V_g, V_{g^{-1}})$ . What can  
you do. Choose a basis for  $V_g$  apply  $\varepsilon$  to get a  
basis for  $\varepsilon V_g = V_{g^{-1}}$ . Then you have a rep relative to  
this basis for both  $\varepsilon$  and  $g$ .

Something happens in the real case which might be interesting. Take  $V = \mathbb{R}^2$   $\varepsilon = \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix}$ . You have  
 $g = F_\varepsilon$  an orthogonal rotation  $g = \begin{pmatrix} \cos \alpha & -\sin \alpha \\ \sin \alpha & \cos \alpha \end{pmatrix}$   $\frac{g+g^{-1}}{2} = \cos \alpha$ .

Start again.  $V$  unitary repn of  $F_\varepsilon$  such that  
 $\frac{g+g^{-1}}{2}$  is scalar operator  $\cos \theta$   $0 < \theta < \pi$ . Then

$V = V_{e^{i\theta}} \oplus V_{e^{-i\theta}}$ . Choose orth. basis for  $V_{e^{i\theta}}$  combine with  
 $\varepsilon^{\text{orth}}$  to get basis for  $V$   $\Rightarrow g = \begin{pmatrix} e^{i\theta} & 0 \\ 0 & e^{-i\theta} \end{pmatrix}$   $\varepsilon = \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix}$

$$\begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} 1 & -1 \\ -1 & 1 \end{pmatrix} = \begin{pmatrix} 1 & -1 \\ -1 & 1 \end{pmatrix} \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix}$$

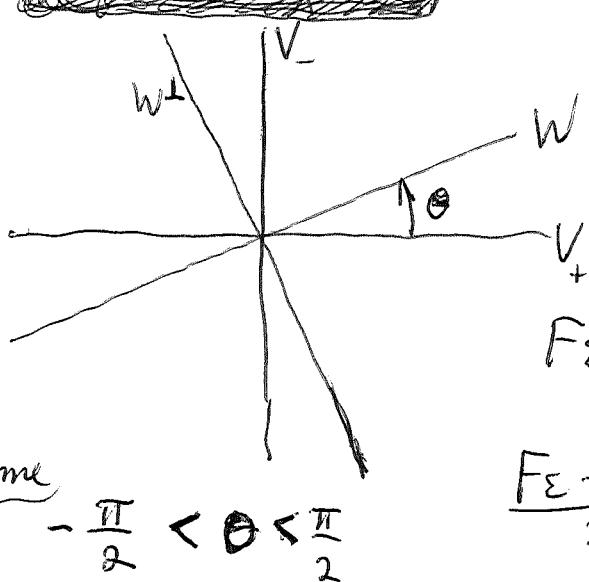
$$\begin{aligned}
 & \frac{1}{2} \begin{pmatrix} 1 & 1 \\ -1 & 1 \end{pmatrix} \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} 1 & -1 \\ 1 & 1 \end{pmatrix} = \frac{1}{2} \begin{pmatrix} 1 & 1 \\ -1 & 1 \end{pmatrix} \begin{pmatrix} 1 & 1 \\ 1 & -1 \end{pmatrix} \\
 & = \boxed{\begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix}} \cdot \frac{1}{2} \begin{pmatrix} 1 & 1 \\ -1 & 1 \end{pmatrix} \begin{pmatrix} e^{i\theta} & 0 \\ 0 & e^{-i\theta} \end{pmatrix} \begin{pmatrix} 1 & -1 \\ 1 & 1 \end{pmatrix} \\
 & = \frac{1}{2} \begin{pmatrix} 1 & 1 \\ -1 & 1 \end{pmatrix} \begin{pmatrix} e^{i\theta} & -e^{i\theta} \\ e^{-i\theta} & e^{-i\theta} \end{pmatrix} = \begin{pmatrix} \cos \theta & -i \sin \theta \\ -i \sin \theta & \cos \theta \end{pmatrix} \quad \text{conj by } \begin{pmatrix} 1 & 0 \\ 0 & i \end{pmatrix}
 \end{aligned}$$

$$\frac{1}{2} \begin{pmatrix} 1 & 1 \\ -i & i \end{pmatrix} \begin{pmatrix} e^{i\theta} & 0 \\ 0 & e^{-i\theta} \end{pmatrix} \begin{pmatrix} 1 & i \\ 1 & -i \end{pmatrix} = \begin{pmatrix} \cos \theta & -\sin \theta \\ \sin \theta & \cos \theta \end{pmatrix}$$

Main question. Given an orthogonal repn of  $F_\Sigma$  on a Euclidean space  $V$  such that  $\frac{1}{2}(g+g^{-1})$  is a scalar operator  $\lambda I$  where  $\lambda \in (-1, 1)$ , can you construct a canonical isom between  $V$  and the 2 diml vred rep  $V$  belonging to  $\lambda \otimes$  vector space.

I think the answer is YES because the endo  
morphism of  $V$  over  $\mathbb{R}$  is  $\mathbb{R}$ .

~~Endomorphisms of  $V$  are represented by  $\mathbb{R}$~~



$$r = \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix} \text{ on } \mathbb{R}^2$$

$$\begin{aligned}
 F \begin{pmatrix} \cos \theta & -\sin \theta \\ \sin \theta & \cos \theta \end{pmatrix} &= \begin{pmatrix} \cos \theta & -\sin \theta \\ \sin \theta & \cos \theta \end{pmatrix} \Sigma \\
 &= \Sigma \begin{pmatrix} \cos \theta & \sin \theta \\ -\sin \theta & \cos \theta \end{pmatrix}
 \end{aligned}$$

$$F_\Sigma = \begin{pmatrix} \cos 2\theta & -\sin 2\theta \\ \sin 2\theta & \cos 2\theta \end{pmatrix}, \quad F = \begin{pmatrix} \cos 2\theta & \sin 2\theta \\ \sin 2\theta & -\cos 2\theta \end{pmatrix}$$

$$\frac{F_\Sigma + \Sigma F}{2} = \cos(2\theta)$$

Assume  $-\frac{\pi}{2} < \theta < \frac{\pi}{2}$

Classify orthogonal irreducible representations  $V$  of the infinite dihedral group  $F, \varepsilon$ . In the group algebra one has the element  $\frac{g+g^{-1}}{2} = \frac{F\varepsilon + \varepsilon F}{2}$  which is symmetric self adjoint. Decompose  $V$  into eigenspaces for  $\frac{g+g^{-1}}{2}$ . Let  $V_K$  be the eigenspace with  $\frac{g+g^{-1}}{2} = K$ .  $V_K \neq 0 \Rightarrow K \in [-1, 1]$ . Look at a  $K$  in  $(-1, 1)$ .

Here's the way to proceed. Take the complex case, that is, unitary reps of  $F, \varepsilon = \langle g \rangle \otimes \langle \varepsilon \rangle$ . s.t.  $\frac{g+g^{-1}}{2} = \lambda$  where  $\lambda \in (-1, 1)$ . Form the quotient ring of  $\mathbb{C}[g, g^{-1}]$  by the relation  $g^2 - 2\lambda g + 1 = 0$ , you get an algebra  $\cong \mathbb{C} \times \mathbb{C}$ . Yes, define  $\mathbb{C}\langle g \rangle \rightarrow \mathbb{C} \times \mathbb{C}$  using the two roots. Then adjoin  $\varepsilon$ , tensoring with  $\mathbb{C}\langle \varepsilon \rangle$  which is 2-dim. Get an alg of dim 4 over  $\mathbb{C}$  quotient of  $\mathbb{C}\langle F, \varepsilon \rangle$ .

$$\boxed{\mathbb{C}\langle F, \varepsilon \rangle / \left( \frac{g+g^{-1}}{2} = \lambda \right)}$$

Real case  $\mathbb{R}\langle F, \varepsilon \rangle \longrightarrow \mathbb{R}\langle F, \varepsilon \rangle / \left( \frac{F\varepsilon + \varepsilon F}{2} = \lambda \right)$

~~$\mathbb{R}\langle g, g^{-1} \rangle$~~   $\mathbb{R}\langle g, g^{-1} \rangle \longrightarrow \mathbb{R}\langle g, g^{-1} \rangle / \frac{g+g^{-1}}{2} = \lambda \xrightarrow{\sim} \mathbb{C}$

Take semi-direct prod.  $\mathbb{C} \rtimes \varepsilon$  dim 4 over  $\mathbb{R}$ .

$$\begin{pmatrix} 1 & 0 \\ 0 & 1 \end{pmatrix} \underbrace{\begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix}}_{-i} \underbrace{\begin{pmatrix} 0 & -1 \\ 1 & 0 \end{pmatrix}}_i \underbrace{\begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix}}_{\varepsilon} \simeq M_4 \mathbb{R}$$

Now what you need is a standard form for these  $F, \varepsilon$  modules satisfying  $\frac{1}{2}(F\varepsilon + \varepsilon F) = \lambda$ . If  $S_\lambda$  is irreducible then since Endo ring of  $S_\lambda$  is  $\mathbb{R}$ , one should have  $\mathbb{R} V \xleftarrow{\sim} S_\lambda \otimes_{\mathbb{R}} \text{Hom}_{F, \varepsilon}(S_\lambda, V)$

$\alpha'$  You want to link this standard picture based on  $S_2$  to what arises from the C.T.

Let  $\lambda \in (-1, 1)$   $\lambda \pm i\sqrt{1-\lambda^2} = \cos \theta \pm i \sin \theta$

$$R_\lambda = R \begin{pmatrix} 1 & 0 \\ 0 & 1 \end{pmatrix} + R \underbrace{\begin{pmatrix} \cos \theta & -\sin \theta \\ \sin \theta & \cos \theta \end{pmatrix}}_g \subset M_2(\mathbb{R})$$

$$\frac{g+g^{-1}}{2} = \cos \theta \begin{pmatrix} 1 & 0 \\ 0 & 1 \end{pmatrix}$$



You want to see that  $R[g, g^{-1}] / (\frac{g+g^{-1}}{2} - \lambda)$  is isom. to  $\mathbb{C}$  via the map  $g \mapsto \lambda + i\sqrt{1-\lambda^2}$

$$R_\lambda = R[g, g^{-1}] / (\frac{g+g^{-1}}{2} - \lambda) \xrightarrow{\sim} \mathbb{C}$$

$\dim 2$

$\varepsilon \xleftarrow{\text{conj. in } \mathbb{C}} \frac{g+g^{-1}}{2} \mapsto \lambda$

$$g \mapsto \lambda + i\sqrt{1-\lambda^2}$$

$$g^{-1} \mapsto \lambda - i\sqrt{1-\lambda^2}$$

$$R_\lambda \times \varepsilon \xrightarrow{\sim} \mathbb{C} \times \varepsilon$$

$a+bi \mapsto \begin{pmatrix} a & -b \\ b & a \end{pmatrix}$

$(a+bi)\varepsilon \mapsto \begin{pmatrix} a & b \\ b & -a \end{pmatrix}$

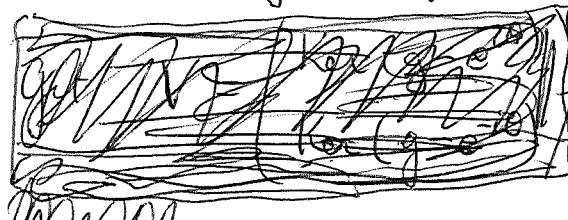
$$\left( \begin{pmatrix} 1 & 0 \\ 0 & 1 \end{pmatrix}, \begin{pmatrix} 0 & -1 \\ 1 & 0 \end{pmatrix}, \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix}, \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \right)$$

?

$\mathbb{C} \otimes_{\mathbb{R}} \mathbb{R}(R + R\varepsilon)$

Idea that since  $-1$  is not an eigenvalue of  $g$  (when  $\lambda \in (-1, 1)$ ), the C.T. theory yields a choice for  $g^{1/2} = \frac{1+x}{(1-x^2)^{1/2}}$

$\beta'$   
 complex case: unitary rep of  $F_E$  on  $V$   
 such that  $\frac{1}{2}(g+g^{-1}) = 1$   $\lambda \in (-1, 1)$  assume.  
~~if g satisfies~~  $g^2 - 2\lambda g + 1 = 0$  eigenvalues  $\lambda \pm i\sqrt{1-\lambda^2}$



~~$V$  splits into eigenspaces for~~

$$V = V_{e^{i\theta}} \oplus V_{e^{-i\theta}}$$

~~If you have to go over~~

$$V = \begin{pmatrix} V_{e^{i\theta}} \\ V_{e^{-i\theta}} \end{pmatrix} = \begin{pmatrix} 1 \\ \varepsilon \end{pmatrix} V_{e^{i\theta}} \quad g = \begin{pmatrix} e^{i\theta} & 0 \\ 0 & e^{-i\theta} \end{pmatrix} \quad \varepsilon = \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix}$$

~~$V = \mathbb{C}^2 \quad g = \begin{pmatrix} e^{i\theta} & 0 \\ 0 & e^{-i\theta} \end{pmatrix} \quad \varepsilon = \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix}$~~

$$\begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} 1 & -1 \\ 1 & 1 \end{pmatrix} \simeq \begin{pmatrix} 1 & -1 \\ -1 & 1 \end{pmatrix} \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix}$$

$$\begin{pmatrix} 1 & i \\ -i & 1 \end{pmatrix} \stackrel{\frac{1}{2}}{\sim} \begin{pmatrix} 1 & 1 \\ -1 & 1 \end{pmatrix} \begin{pmatrix} e^{i\theta} & 0 \\ 0 & e^{-i\theta} \end{pmatrix} \begin{pmatrix} 1 & -1 \\ 1 & 1 \end{pmatrix} \sim \begin{pmatrix} 1 & 0 \\ 0 & 1 \end{pmatrix}$$

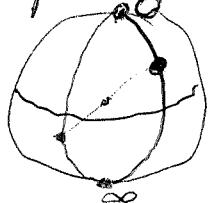
$$\frac{1}{2} \begin{pmatrix} 1 & 1 \\ -1 & 1 \end{pmatrix} \begin{pmatrix} e^{i\theta} & -e^{i\theta} \\ e^{-i\theta} & e^{-i\theta} \end{pmatrix} = \begin{pmatrix} \cos \theta & -\sin \theta \\ \sin \theta & \cos \theta \end{pmatrix}$$

$$\frac{-e^{i\theta} + e^{-i\theta}}{2}$$

$$\frac{1}{2} \begin{pmatrix} 1 & 1 \\ -i & i \end{pmatrix} \begin{pmatrix} e^{i\theta} & 0 \\ 0 & e^{-i\theta} \end{pmatrix} \begin{pmatrix} 1 & i \\ 1 & -i \end{pmatrix}$$

$$-i \sin \theta$$

Riemann sphere  $P^1(\mathbb{C})$  An involution on  $\mathbb{C}^2$   
 corresponds to (is the same as) an ordered pair of  
 antipodal points.  $z \mapsto -\bar{z}^{-1}$   $\frac{1}{t} \mapsto -\bar{t}$



So you see there is a phase defined involving the longitudes.

Next you want to look at the other picture of a unitary rep of  $F, \varepsilon$  such that  $(g+g^{-1})/2 = \lambda \in (-1, 1)$ . Namely, eigenspaces

$$V_{\xi} = \text{Ker}(g-\xi), \quad V_{\bar{\xi}} = \text{Ker}(g-\bar{\xi})$$

$$\xi = \lambda + i\sqrt{1-\lambda^2} = e^{i\theta}$$

$$V = \begin{pmatrix} V_{\xi} \\ V_{\bar{\xi}} \end{pmatrix}$$

$$g = \begin{pmatrix} \xi & 0 \\ 0 & \bar{\xi} \end{pmatrix} \quad \varepsilon = \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \quad F = \begin{pmatrix} 0 & \xi \\ \bar{\xi} & 0 \end{pmatrix}$$

~~What's next?~~ Next you want to convert  $\varepsilon$  to the standard form  $\begin{pmatrix} 0 & 0 \\ 0 & -1 \end{pmatrix}$  (corresponding to ~~electric~~ magnetic)

$$\frac{1}{2} \begin{pmatrix} 1 & 1 \\ -i & i \end{pmatrix} \begin{pmatrix} e^{i\theta} & 0 \\ 0 & e^{-i\theta} \end{pmatrix} \begin{pmatrix} 1 & -i \\ 1 & -i \end{pmatrix} = \begin{pmatrix} \cos\theta & -\sin\theta \\ \sin\theta & \cos\theta \end{pmatrix}$$

You can further conjugate by  $\begin{pmatrix} 1 & 0 \\ 0 & e^{i\varphi} \end{pmatrix}$ , which fixes ~~the standard~~  $\varepsilon = \begin{pmatrix} 0 & 0 \\ 0 & -1 \end{pmatrix}$ .

For the model  $g = \frac{1+X}{1-X}$   $X = \begin{pmatrix} 1 & -T^* \\ T & 1 \end{pmatrix}$

Riemann sphere

~~$\varepsilon \tau = \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix} z = \frac{z}{-1} = -z$~~

$$e^{2i\theta} z$$

~~$\varepsilon \tau = \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix} z = \frac{z}{-1} = -z$~~

$$\begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} z = \frac{1}{z}$$

$$\begin{pmatrix} e^{i\theta} & 0 \\ 0 & e^{-i\theta} \end{pmatrix} (z) = \frac{e^{i\theta} z + 0}{0 + e^{-i\theta} z}$$

~~so~~

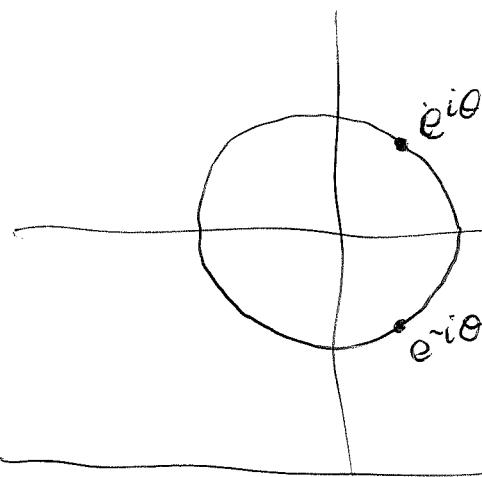
any involution fixes 2 pts.

$$\begin{pmatrix} e^{i\theta} & 0 \\ 0 & e^{-i\theta} \end{pmatrix} \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix}$$

$$= \begin{pmatrix} 0 & e^{i\theta} \\ e^{-i\theta} & 0 \end{pmatrix}$$

$$\delta' \quad \begin{pmatrix} 0 & e^{i\theta} \\ e^{-i\theta} & 0 \end{pmatrix}(z) = \frac{0z + e^{i\theta}}{e^{-i\theta}z + 0} = e^{2i\theta}z^{-1}$$

fixpts are  $e^{2i\theta} = z^2 \Rightarrow z = e^{\pm i\theta}$

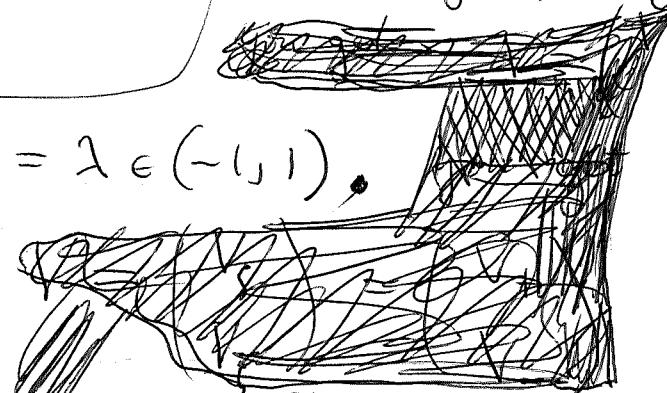


$$g z = \cancel{e^{2i\theta}} z$$

$$\begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} z = \cancel{e^{2i\theta}} z^{-1}$$

Review the problem. Given a unitary repn of  $F_\mathbb{C}$  on  $V$  such that

$$\cancel{g} \frac{1}{2}(g + g^{-1}) = \lambda \in (-1, 1),$$



You want ~~to~~ an isomorphism of representations between  $V$  and ~~standard~~ standard mixed repn ~~(it's)~~  $\otimes$  vectorspace.

~~to~~ You have two methods to ~~construct~~ construct  $V \cong U_\lambda \otimes ?$

What is your aim? You start with a unitary repn on  $F_\mathbb{C}$  on a Hilbert space  $V$  such that  $\frac{1}{2}(g + g^{-1}) = \lambda \in (-1, 1)$ . You have ~~standard~~ an irreducible 2 diml repn  $U_\lambda$  with these properties. You want to construct a canonical isomorphism  $V = U_\lambda \otimes \text{Hom}_{F_\mathbb{C}}(U_\lambda, V)$  ~~respecting~~ respecting herm. scalar product

At this point I am beginning to believe  
 that ~~the~~ the <sup>vague</sup> idea about coherent phases is  
 naives explain, describe idea. You consider a  
 unitary (f.d.) repn of  $F, \varepsilon$  such that  $g = F\varepsilon$  sat  
 $\frac{1}{2}(g + g^{-1}) = \lambda \in (-1, 1)$ . You seek a canonical picture  
 of this rep. (IDEA: Every automorphism of a simple  
 (maybe semi-simple) <sup>f.d.</sup> algebra is inner. So such an autom.  
 determines an invertible element modulo ~~a~~ a central  
 invertible. In the real case this should be  $\pm 1$ , in the  
 (case  $S^1$ ))

If you have any hope of utilizing ~~the~~ the  
 phases ignored (or suppressed) by taking the  
 polar decomposition of the classical motion operator,  
 then you need somehow to get the correct ~~the~~  
 (i.e. Bose-Einstein) statistics. There might be  
 possibilities of ~~the~~ utilizing algebraic K-theory  
 ideas.

Basic Idea:  $A_\lambda = C^*$ -algebra gen. by  $F, \varepsilon \rightarrow \frac{1}{2}(g + g^{-1}) = \lambda$   
 where  $\lambda$  assumed in  $(-1, 1)$ .  $A_\lambda$  should be isomorphic  
 to  $M_2(\mathbb{C})$  as  $C^*$  alg. You have a \*-rep  $A_\lambda \rightarrow M_2(\mathbb{C})$   
 s.t. ~~isomorphism~~  $g \mapsto \begin{pmatrix} e^{i\theta} & 0 \\ 0 & e^{-i\theta} \end{pmatrix}$ ,  $\varepsilon \mapsto \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix}$

where ~~is~~  $\cos \theta = \lambda$ . This \*-rep should be a  $C^*$ -alg ~~isom.~~ <sup>V<sub>1</sub></sup>  
 You should know that  $A_\lambda$  has a unique irreducible \*-repn.  
 up to isomorphism. ~~isom.~~ The isom should be unique up to  
~~isomorphism of~~ <sup>V<sub>1</sub></sup> an autom. of  $V_1$ , and such an autom is  
 mult by a scalar in  $S^1$ .

~~isom.~~ A arbitrary (say f.d.) \*-repn. of  $A_\lambda$  should  
 be isomorphic to  $V_1 \otimes E$ , where  $E$  is a f.d. Hilb space.

~~isom.~~ If you restrict to reps of mult.  $n$ , the arbitrariness  
 in the choice of  $E$  should amount to the space  $U_n$ .

Real case  $A_{\lambda, \mathbb{R}}$  = real  $C^*$ -alg with same gens. and relations. You should get an irred \*repsn. on  $\mathbb{R}^2$  given by  $g \mapsto \begin{pmatrix} \cos \theta & -\sin \theta \\ \sin \theta & \cos \theta \end{pmatrix}$   $\epsilon \mapsto \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix}$ . The centralizer of this repsn., call it  $V_{\lambda, \mathbb{R}}$ , should be  $\mathbb{R}$ , so the autos respecting the  $\| \cdot \|$  should be  $\pm 1$ . Then ~~there are~~ an arbitrary ~~choice~~ \*repsn of  $A_{\lambda, \mathbb{R}}$  should have the form  $V_{\lambda, \mathbb{R}} \otimes E$   $E$  Euclidean space. If  $E$  has dim  $n$ , then the arbitrariness in the choice of  $E$  should amount to  $O_n$ .

How does this compare to the C.T. picture. In the complex case you have  $g = \frac{1+x}{1-x}$  where ~~there are~~  $x = \begin{pmatrix} 0 & -T^* \\ T & 0 \end{pmatrix}$  where  $T: V_+ \rightarrow V_-$  satisfies  $TT^* = T^*T = \mu^2$

$$\frac{g+g^{-1}}{2} = \frac{1+x^2}{1-x^2} = \frac{1-\mu^2}{1+\mu^2} = \cos \theta$$

$$\text{so } \mu = \left| \tan \left( \frac{\theta}{2} \right) \right|$$

$$\begin{array}{c} \boxed{\begin{matrix} V_+ & \\ 0 & \end{matrix}} \\ \downarrow \\ \boxed{\begin{matrix} 1 & \\ 0 & \end{matrix}} \end{array} \mu \quad \cos \left( \frac{\theta}{2} \right)^2 = \frac{1}{1+\mu^2}$$

$$\sin \left( \frac{\theta}{2} \right)^2 = \frac{\mu^2}{1+\mu^2}$$

What do you want? You seek a standard picture, canonical form, for an ~~arbitrary~~ orthogonal repsn of  $F, E$  such that  $\frac{1}{2}(g+g^{-1}) = \lambda \in (-1, 1)$ . ~~So that all do you want?~~

From the C.T. viewpoint you get

$$\begin{aligned} g &= \begin{pmatrix} 1 & -T^* \\ T & 1 \end{pmatrix} \begin{pmatrix} 1+T^*T & 0 \\ 0 & 1+T^*T \end{pmatrix}^{-1} \\ &= \begin{pmatrix} 1-TT^* & -2T^* \\ 2T & 1-TT^* \end{pmatrix} \begin{pmatrix} 1+T^*T & 0 \\ 0 & 1+T^*T \end{pmatrix}^{-1} \\ &= \begin{pmatrix} \cos \theta & (-\sin \theta)u^* \\ (\sin \theta)u & \cos \theta \end{pmatrix} \end{aligned}$$

$$\begin{aligned} \sin \theta &= \\ 2 \sin \frac{\theta}{2} \cos \frac{\theta}{2} &= \\ \frac{2\mu}{1+\mu^2} & \end{aligned}$$

$$u: V_+ \xrightarrow{\sim} V_-$$

~~orthogonal~~

$\eta'$  Look at harmonic oscillator from the general symplectic viewpoint. What you've been looking at is harmonic oscillators from LC circuits. Review the ~~mechanics~~ mechanics situation  $L = T - U = \frac{1}{2} \dot{x}^T m \dot{x} - \frac{1}{2} \dot{x}^T k x$ ,  $p = \frac{\partial L}{\partial \dot{x}} = m \dot{x}$   $H = \dot{p} \dot{x} - L(x, \dot{x})$  but with  $\dot{x}$  put =  $\vec{p}$  then  $H = p^T m^{-1} p - \frac{1}{2} (m^{-1} p)^T m (m^{-1} p) + \frac{1}{2} x^T k x$   
 $= \frac{1}{2} p^T m^{-1} p + \frac{1}{2} x^T k x$   
 $\dot{x} = \frac{\partial H}{\partial p} = m^{-1} p$   $\dot{p} = -\frac{\partial H}{\partial x} = -k x$   
 $(\dot{x})^* = \begin{pmatrix} 0 & m^{-1} \\ -k & 0 \end{pmatrix} (\dot{x})$  This matrix should be skew-symmetric wrt  $H$ .

~~Hamilton's principle tells us that you have~~ It's clear that you have a polarized Euclidean space corresponding to the two energy types: kinetic + potential. Things are mysterious <sup>from the</sup> Legendre transform ~~to~~ to Hamilton's equations.  
 There should be a good viewpoint, probably using path integrals. ~~Hamilton's principle~~  $\delta \int L dt = 0$  which yields  $p = \frac{\partial L}{\partial \dot{x}}$  and the contact transformation between  $t=a$  and  $b$ :  $[p \delta q]_a^b = 0$

How do you ~~handle~~ handle LC ~~networks~~ networks?  
~~An easier question~~ An easier question might be to take the symplectic ~~situation~~ situation and ask ~~if it fits into~~ if it fits into Hamilton's equations. Recall the basic idea that  $\Gamma_s = \begin{pmatrix} s & 0 \\ 0 & s^{-1} \end{pmatrix} \begin{pmatrix} \{V_C\} \\ \{I_L\} \end{pmatrix} \subset \begin{pmatrix} V_+ \\ V_- \end{pmatrix}$  should be transversal to the Kirchhoff ~~editon~~ space.

$$\theta' \quad \begin{pmatrix} x \\ p \end{pmatrix}' = \underbrace{\begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix}}_{\text{symplectic form}} \underbrace{\begin{pmatrix} k & 0 \\ 0 & m^{-1} \end{pmatrix}}_{\text{Hamiltonian quad form}} \begin{pmatrix} x \\ p \end{pmatrix}$$

tomorrow (after lecture preparation!) you need to  
~~not~~ go over again the flow  $X$  arising from  
 $\omega$  and  $H$ .

The problem is to start with  $H$ , i.e. treat phase space as a Euclidean space equipped with a skew-symmetric non ~~singular~~<sup>singular</sup> transformation  $X$ . The phase of  $X$  ~~gives~~ gives a complex structure. ??

~~Recall~~ It should be clear 3 objects on phase space  $\mathcal{Q}$ :

- 1) symplectic form
- 2) Hamiltonian  $H$  (pos. def.)
- 3) ~~invertible~~ invertible operator  $X$

Roughly two of these  $\Rightarrow$  third  
 (with appropriate compatibility)

Recall

$$\begin{pmatrix} s & T^* \\ -T & s \end{pmatrix} \begin{pmatrix} \hat{V}_C \\ \hat{I}_L \end{pmatrix} = \begin{pmatrix} V_C(0) \\ I_L(0) \end{pmatrix}$$

IVP for LC network.

$$s-X \text{ where } X = \begin{pmatrix} 0 & -T^* \\ T & 0 \end{pmatrix}$$

~~Therefore~~ Therefore ~~an~~ an LC network ~~yields~~ seems to a harmonic oscillator of the ~~phase space~~ phase space, kinetic + potential energy types.  $X$  is skew-symmetric ~~is~~ invertible on  $V = \begin{pmatrix} V_+ \\ V_- \end{pmatrix}$ . Polar decomposition for  $X$  and  $T: V_+ \rightarrow V_-$  should be equivalent?

i' Start with a real vector space  $V$  equipped with a symplectic bilinear form and a positive def symmetric form, whence  $V$  is a Euclidean space equipped with an invertible skew-symmetric operator  $X$ . Then apply spectral theory to  $X$ . Just what is spectral theory of  $X$ ? Say  $\dim V=2$ . You pick an orth basis for  $V$ , the matrix for  $X$  should be  $\begin{pmatrix} 0 & -a \\ a & 0 \end{pmatrix}$ ,  $a \neq 0 \in \mathbb{R}$ . So frequency is  $|a|$  and the ~~positive~~ direction of rotation is given by the sign of  $a$ .

**Vague Idea:** Extract a square root of a negative quantity. There's an ambiguity in sign which you might treat as a random variable.

~~Suppose that your oscillator has~~ Suppose that your oscillator has a single frequency, that is,  $X^2 = -\omega^2$ , (note  $\delta^2 = -\omega^2$ !)

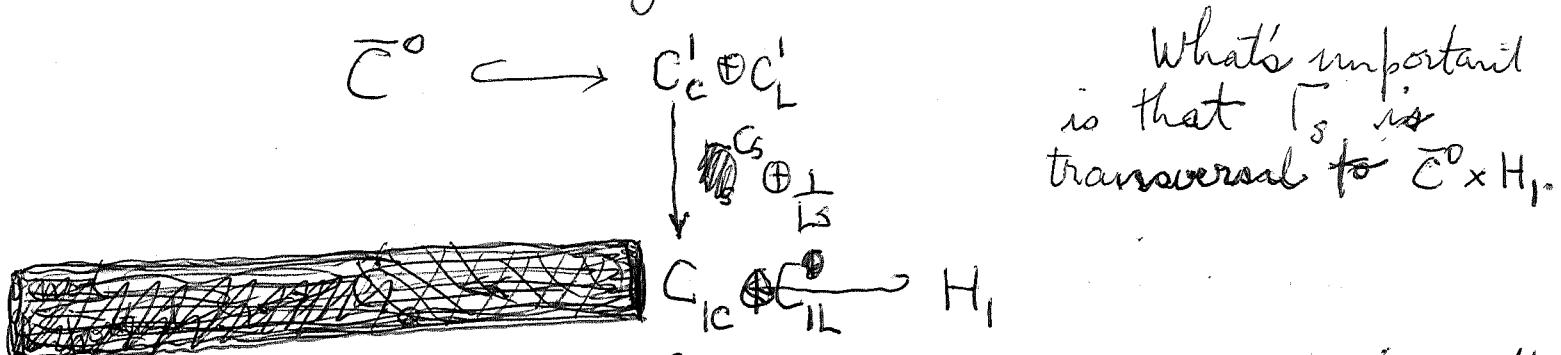
$O(2n)/U(n) =$  space of complex structures on the Euclidean space  $\mathbb{R}^{2n}$ . These are the possible square roots of  $-I$ .

Recall your old old ideas about buildings using finite  $\dim$  unitary groups + self adjoint  $A$ ,  $0 \leq A \leq I$ . There should be some connection with ~~less old~~ work on the moment map. (Atiyah, Kirwan, Guillimin-Stern)

Complex analog, consider f.d Hilbert space and a non-deg skew-hermitian operator  $X$ . Then  $X^*X = -X^2$  so  $|X| = (-X^2)^{1/2}$  and  $J = X|X|^{-1}$  satisfies  $J^2 = X^2|X|^{-2} = X^2(-X^2)^{-1} = -I$ . Then you can identify  $J$  with the involution  $F$  given by  $iF = J$ . In other words, because  $i$  is already present, an operator  $J$  satisfying  $J^2 = -I$  is equivalent to the ~~the~~  $F$  ~~other~~ corresponding to the eigenspaces of  $J$ . Except you have to pick either  $i$  or  $-i$  for  $F = \pm I$ .

K' So in the complex case the space of skew-hermitian  $X$  on  $\mathbb{C}^n$  satisfying  $X^2 = -\omega^2$  ( $\omega \neq 0$ ) is the space of involutions  $\{F \mid F = F^*, F^2 = 1\}$ . This is the total Grassmannian  $\coprod_{0 \leq p \leq n} \text{Gr}_p(\mathbb{C}^n) : \frac{\mathcal{U}(n)}{\mathcal{U}(n) \times \mathcal{U}(p)}$

Idea not to forget, about an LC network with "external" nodes + ~~the corresponding response function.~~ You probably also want a systematic treatment of ~~a~~ forced harmonic oscillator. There is probably ~~a~~ quotient symplectic picture. ~~the~~

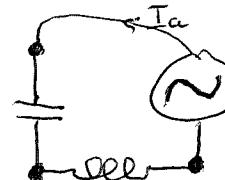
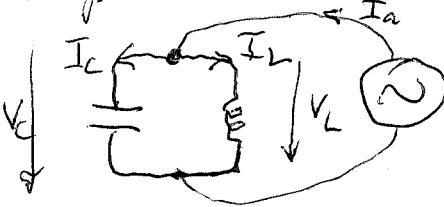


When you label 2 nodes in the graph "external" you fix the voltage drop between these nodes, which means that you restrict the node voltage function to lie in a codim 1 coset (affine hyperplane in  $C^\circ$ ). So you decrease  $C^\circ$ . The voltage drop across the external nodes should be viewed as an applied voltage source which causes a current to flow between the external nodes.

~~the network~~ In effect you are changing the network by adding an edge joining the two external nodes and putting <sup>the applied</sup> voltage source in this edge. You're dealing with the Thevenin equivalent circuit. ?

You are trying to understand the inhomogeneous equation, i.e. the motion of an LC network with an applied voltage source. Recall that you have ~~the~~ first order ODE describing the motion in which <sup>the</sup> 2e dynamical variables  $(V_C, I_o)$  ~~the~~ reduce to the ~~the~~ dominant variables  $V_C, I_L$  modulo the Kirchhoff constraints. This ~~means~~ <sup>should</sup> that once the forcing terms are expressed in terms of  $V_C, I_L$  then ~~the~~ the solution of the perturbed motion is clear.

~~What~~ What might be a good question? Look at simple harmonic oscillator. 1 node other than the ground.



$$V_a = V_a + V_L = \left( Cs + \frac{1}{Ls} \right) I_a$$

Problem: ~~When~~ When you connect an applied voltage source does this change the homogeneous situation? That is, the dominant variables.

$$Z = \frac{V_a}{I_a} = \frac{1}{Cs + \frac{1}{Ls}} = \frac{Ls}{Ls^2 + 1}$$

$$V_C = V_L = V_a$$

$$I_a = I_C + I_L = CsV_C + \frac{1}{Ls}V_L = \left( Cs + \frac{1}{Ls} \right) V_a$$

pulling back

You are bothered by a memory of ~~reducing~~ a quadratic form ~~to~~ to a subspace, and pushing it down to a quotient space. ~~You spent a lot of time on this~~ You spent a lot of time on this ~~idea~~ without getting the free motion of the LC network.

$\mu'$  Quadratic extension of  $\mathbb{Q}$ . ~~is~~  $\mathbb{Q}(\sqrt{a}) = \mathbb{Q} + \mathbb{Q}\sqrt{a}$

$$= \{ \alpha + \beta\sqrt{a} \mid \alpha, \beta \in \mathbb{Q} \}.$$

~~subset~~

~~of  $\mathbb{Q}$~~   ~~$a \in \mathbb{Q}$~~   ~~$\sqrt{a}$~~

quadratic equation over  $\mathbb{Q}$ .

$$x^2 + bx + c = 0 \quad x = -b \pm \sqrt{b^2 - c}$$

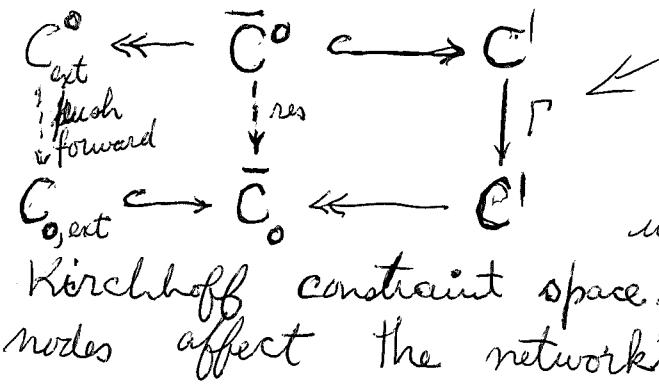
Let  $a \in \mathbb{Q}$  assume  $b \in \mathbb{Q}$  st.  $b^2 = a$ .

back to harmonic oscillators. For some reason you thought that ~~an~~ an applied voltage at some node produced a response function.

Explain: Begin with a connected LC network. You want to apply a voltage source ~~between~~ <sup>between</sup> some nodes to the ground. Thevenin idea should explain what's happening.

The real puzzle is why you were ~~able~~ able to get a response function at an external node without understand the free oscillations. ~~but~~ Actually this is not surprising, because it's easy to find the steady state motion arising a forcing term with a fixed frequency  $\omega$ , not one of the normal modes.

What's puzzling to me is how to connect, link your result [about a subquotient of a polarized Euclidean space being equivalent to a type of response function] to your ~~picture~~ picture of the free motion.



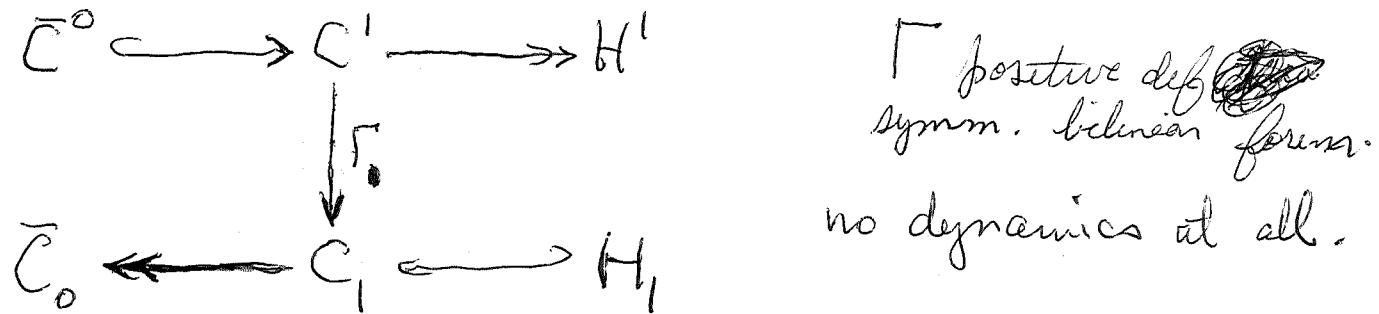
What's wrong here is that there is nothing about the other side, the loop current space  $H$ , which together with  $C^0$  is the Kirchhoff constraint space.

Problem: How do the external nodes affect the network? You apply a voltage at the

$V'$  node which fixes the potential at this node, and you get a current passing through the node. You need to analyze restricting ~~the~~ voltage function ~~to~~ and also allowing a node current. What ~~is~~ does this do to the space of loop currents?  $\square$

So what should be the approach to an applied voltage? Ideas. There should be a dynamical ~~response~~ response not confined ~~to~~  $C^0$ . Something involving <sup>node</sup> currents  $\xrightarrow{\text{The node voltage space}}$  might you expect that declaring a node to be external ~~should~~ change the graph by adding an extra edge joining the external node to ground.

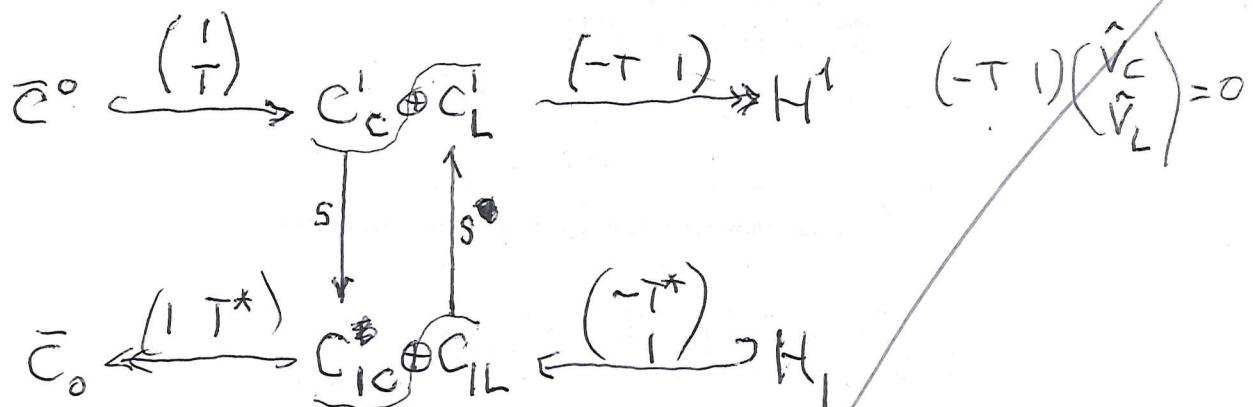
Ideas: Thevenin, look at resistance network response



Suppose you ~~make~~ make one of the nodes ( $\neq$  ground) external, and apply a voltage source. Two changes: the potential at the <sup>external</sup> node is fixed, there is a <sup>possible</sup> node current. It looks like  $H_1$  increases and  $C^0$  decreases.

Restricting <sup>(the)</sup> node potential to a hyperplane should be an inhomogeneous constraint, which is expressed via a map  $C^1 \rightarrow H^1 \times \mathbb{R}$ . Maybe this is the same as adding an edge joining the ground to the external node.

9' Go back to ~~the other~~ the IVP  
for a closed LC network.



$$\dot{V}_C = I_C \quad \dot{I}_L = V_L$$

$$L \dot{V}_C = -V_C(0) + sL V_C$$

$$\hat{V}_L = \hat{I}_L = -I_L(0) + s \hat{I}_L \quad \text{and} \quad T \hat{V}_C = \hat{V}_L$$

$$\hat{I}_C = \hat{V}_C = -V_C(0) + s \hat{V}_C \quad \hat{I}_C = -T^* \hat{I}_L$$

$$s \hat{V}_C + T^* \hat{I}_L = V_C(0)$$

$$-T \hat{V}_C + s \hat{I}_L = I_L(0)$$

$$\left\{ s - \begin{pmatrix} 1 & -T^* \\ T & 1 \end{pmatrix} \right\} \begin{pmatrix} \hat{V}_C \\ \hat{I}_L \end{pmatrix} = \begin{pmatrix} V_C(0) \\ I_L(0) \end{pmatrix}$$

There should be no problem about the forced oscillator here. This is a first-order ordinary linear DE constant coeff. The forcing term can be any function of time with values in the space of dominant variables. This means that you can put a varying voltage source

in series with a capacitor for each capacitance edge, and a varying current source in series with the inductor for each inductance edge.

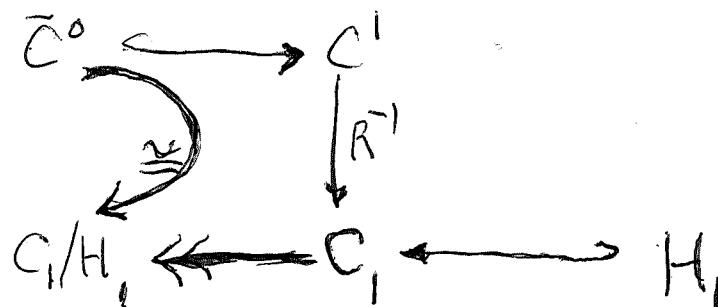
This seems to be the most you can say about a given connected LC network.

Next you want to apply a voltage to a node (really: from the ground to the node). This means you add an edge to the graph, where the edge contains the voltage source.

The interesting point is that the edge has neither L nor C type. If the

Recall the idea that in a resistance network it is useful to assume each edge is a pure e.m.f. in series with an internal resistance. (This is part of the Thevenin thm.)

Review R networks.



DeRho ex.



The problem: How to handle no resistance of the attached edge. No problem because if your voltage source is a battery, its internal resistance is zero.

You have to handle the case where the resistance of the new edge is 0, maybe also the case  $R = \infty$ . The point should be that even though the Ohm's Law correspondence  $\Gamma$

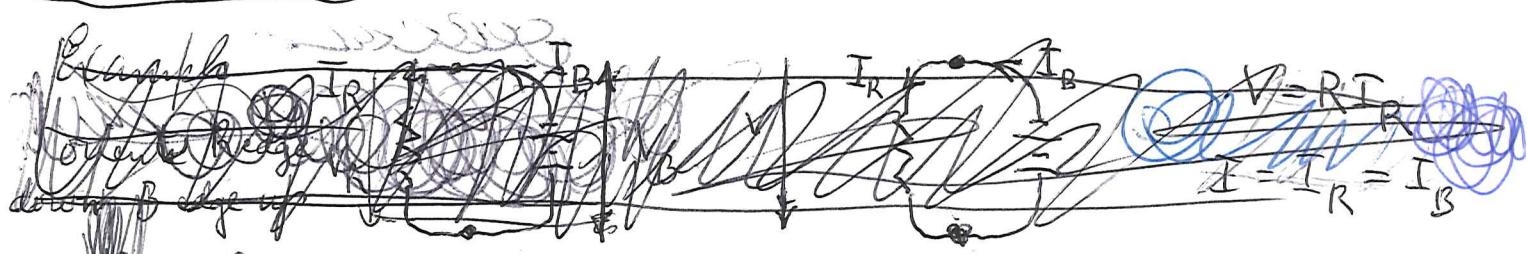
$\pi'$  between  $C'$  and  $C_1$  is singular, it might be nonsingular modulo the Kirchhoff space, i.e.  $\Gamma$  is a complement for  $\bar{C}^o \oplus H_1$  in  $C' \oplus C_1$ .

Let  $\sigma$  denote the added edge, let  $\bar{C}^o, C_1, H_1$  pertain to the original graph. Then the augmented graph has extra variables  $V_\sigma, I_\sigma$ . Its picture

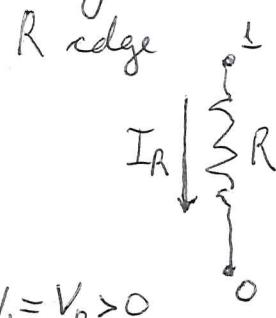
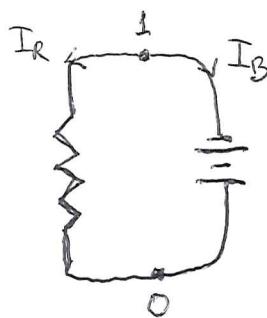
$$\bar{C}^o \hookrightarrow C' \oplus \{V_\sigma\} \longrightarrow H_1 \oplus \{I_\sigma\}$$

$$\bar{C}_o \leftarrow C_1 \oplus \{I_\sigma\} \leftarrow H_1 \oplus \{I_\sigma\}$$

It's probably better to begin with ~~a conn.~~ a resistance network and remove an edge. ~~such that the new graph is connected.~~



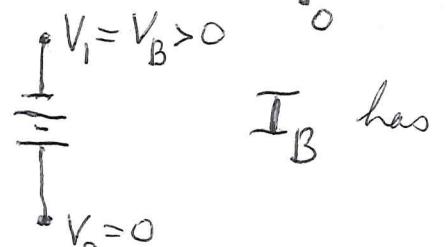
Orient edges by ordering the nodes



The positive direction is from 1 to 0. To

$$V_R = V_1 - V_0$$

B edge direction

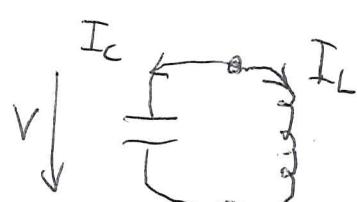


$I_B$  has wrong sign

$$I_B = -T_R$$

$$T R_C = V_L$$

$$I_C + T^* I_L = 0$$



$$\bar{C}^o \xrightarrow{\text{+}} C' \xrightarrow{\text{-}} H_1$$

$$\bar{C}_o \xrightarrow{\text{(1) } T^*} C_1 \xrightarrow{\text{(-)}} H_1$$

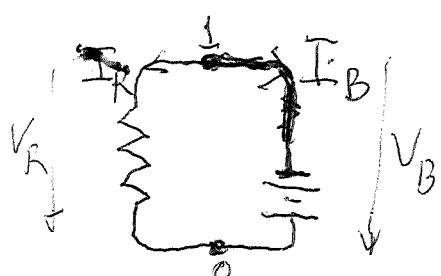
$$s \hat{V}_C - \hat{I}_C = V_C(0)$$

$$-T \hat{V}_L + s \hat{I}_L = I_L(0)$$

$$\begin{pmatrix} s & T^* \\ -T & s \end{pmatrix}$$

S' external edge joining different nodes  
 O resistance pure emf source. What equations  
 do you need to solve for the response?  
 First suppose the external edge has a (small) resistance.  
 The equations to solve should be inhomogeneous,  
 the associate homogeneous ~~equations~~ equations are  
 Ohm's Law for the edges (e equations) ~~plus~~ together  
 with Kirchhoff's 1st + 2nd Laws (e equations).  
 The inhomogeneous term ~~will arise from~~ the applied emf on  
 the external edge.

What form should the "forcing" term take?  
 Recall that the full collection of  $2e$  state  
 variables is redundant because of the  $e$  Kirchhoff  
 constraints. So you probably have to select  $e$   
 state variables which are independent of ~~the~~ the  
 Kirchhoff constraints, like choosing the dominant  
 variables in the LC case. Back to



$$\{V_1\} \xrightarrow{(1)} \begin{pmatrix} V_R \\ V_B \end{pmatrix} = \begin{pmatrix} V_1 \\ V_1 \end{pmatrix} \xrightarrow{(1-1)} \{ \}$$

$$(R \ 0) \quad ??$$

$$V_R = V_B = V_1$$

$$\{I_1\} \xleftarrow{(1)} \begin{pmatrix} I_R \\ I_B \end{pmatrix} \xleftarrow{(-1)} \{I_2\}$$

$$I_R = I_B = -I_B$$

$$V_B \text{ fixed} = V_1$$

$$V_R = R I_R$$

Let's try again to understand the 2 edge loop circuit. There are 4 variables  $V_R, I_R, V_B, I_B$ . The arrows on the edges indicate the orientation chosen for these edges.  $I_R$  and  $I_B$  are expected to be positive.  $V_R$  is a positive voltage drop, starting at the beginning at the top node is higher than the ending at the bottom node. But  $V_B$  is negative because it starts with the potential zero at the ground, and it ends with the higher potential at the top node.

You have the <sup>these</sup> Kirchhoff Ohm equations

$$V_R + V_B = 0, \quad I_R = I_B, \quad V_R = R I_R$$

There does not seem to be a 4th equation linking  $I_B, V_B$ .

But the three equations yield the relation

$$-V_B = +V_R = +R I_R = +R I_B$$

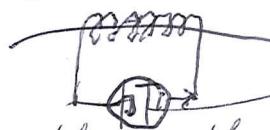
between the voltage drop and current in a battery edge.

The situation is unsatisfactory. You hoped to get a clear picture of the effect of applied voltage and applied current sources, by exploiting Thevenin theory. In this theory one treats the applied source by adding a new edge to the graph.

Idea: Voltage source is attached to nodes; it involves  $\bar{C}^o$ , really a linear functional on  $\bar{C}^o$ , an element of  $\bar{C}_o$ . So maybe an applied current source

5' should be associated to a linear functional on the space  $H_1$ , i.e. an element of  $H^*$ . You would then be linking ~~external~~ external sources to the dual of the Kirchhoff space.

Here's how to produce a current source in an edge, namely, put the wire through a solenoid



A.C. Voltage source

run current thru the solenoid, you get a magnetic field induced in the wire which should induce a alternating current in the wire. ~~Maybe simpler would be to an A.C. generator with permanent magnet field core.~~

The central problem is to understand forcing terms, inhomogeneous terms in an LC network. There is a straightforward answer when you start with the IVP for the network.

But what happens when you are not given the forcing term in terms of the dominant variables?

Ideas look at s.h.o., see if you can handle non dominant forcing terms.

Elec. Eng State Variables  $\begin{pmatrix} A & B \\ C & D \end{pmatrix}$ , somehow this yields a  $C(s-A)^{-1}B + (D?)$ . Something similar occurs when you have a space of 1-particle states for a quantum field, the Hamiltonian is  $A$ , interacting - being scattered by heavy particle

$$|A| + |B| \geq |A+B|$$

To what to do next? Maybe look at the degenerate frequencies. Start with polarized Euclidean space and subspaces. Everything gets split into s.h.o.'s and certain degenerate cases. Exactly what should be related to  $g = \pm 1$ , so the B,C wings should perhaps be linked to L and C. Question about  $g = \pm 1$  and frequencies 0,  $\infty$ ?

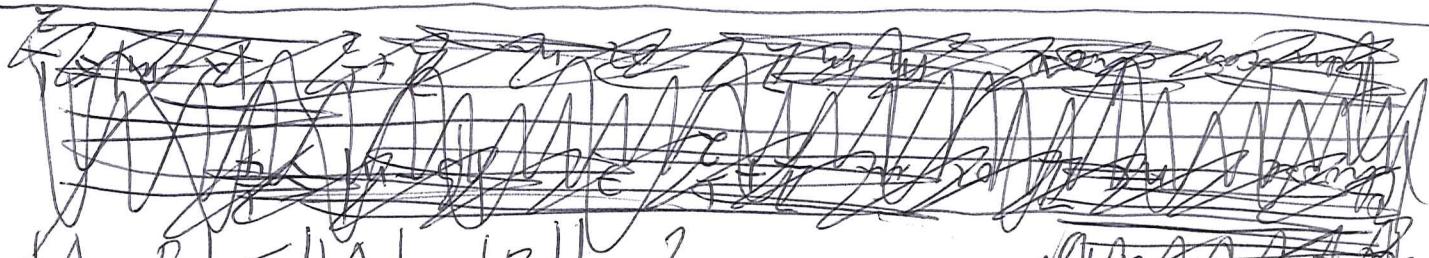
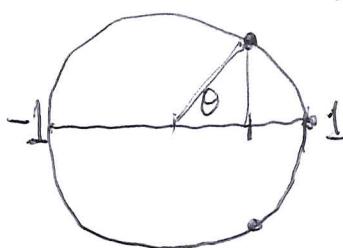
Idea: Eliashberg ~~rigidity~~ then: a  $C^\infty$  limit of  $C^\infty$  symplectic transformations ~~is~~ is a  $C^\infty$  limit. Better: a sequence of  $C^\infty$  symplectic transformations which ~~converges~~ converges in the  $C^\infty$ -topology in fact converges in the  $C^\infty$ -topology. You might be able to use the  $C^\infty$  limit to prove decays

Return to the Grassmannian  $SL_2(\mathbb{R}) \times \text{stable} \rightarrow \text{closed}$   $U(1)P$

You have polarized Euclidean space  $V = \begin{pmatrix} V_+ \\ V_- \end{pmatrix}$  and a subspace  $W \subset V$ , whence  $F = \pm 1$  on  $W$  ( $\pm 1$  on  $W^\perp$ )  $\varepsilon = \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix}$  on  $W$  and other then decompose according to spectrum of  $\frac{1}{2}(g+g^{-1})$ ,  $g=F\varepsilon$ . You should split  $V, F, \varepsilon$  into s.h.o.'s for eigenvalues  $\cos \theta \in (-1, 1)$ .

Next you need to understand the cases

$$\begin{array}{ll} g = +1 & \text{whence } F = \varepsilon \\ g = -1 & \text{--- } F = -\varepsilon \end{array}$$



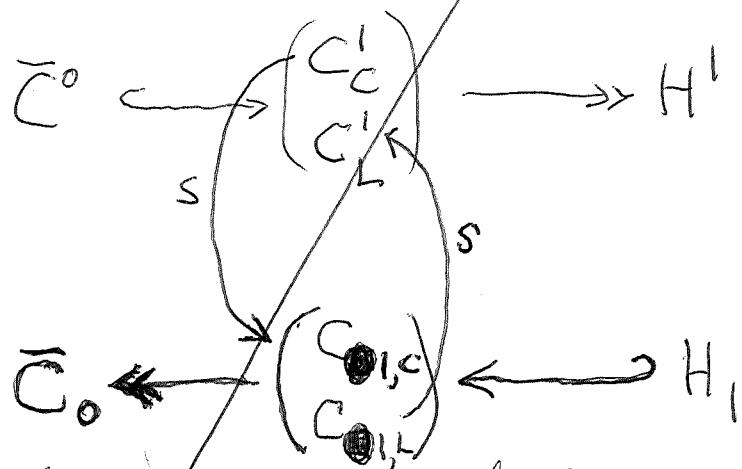
$$|A - B| \leq ||A| - |B|| ?$$

4'

Back to  $V = (V_+, V_-)$ .  $W \subset V$ , you've been looking at the case where  $W = (\frac{1}{T})V_+$  with  $T: V_+ \rightarrow V_-$ . This occurs when  $W \rightarrow V_+$  is an isomorphism.

~~that's not the case~~ In general it has a kernel + cokernel - kernel =  $W \cap V_-$ , cokernel  $W^\perp \cap V_+$ .  $W \cap V_-$  is where  $F=+1, \varepsilon=-1$  and  $W^\perp \cap V_+$  is where  $F=-1$  and  $\varepsilon=+1$ . These comprise the -1 eigenspace for g. ~~Observe also,~~ ~~this part of the proof is not yet complete~~ Remove these, reduce to case where  $W \cong V_+$  whence  $W = (\frac{1}{T})V_+$ , then look at proj.  $W \rightarrow V_-$ . Kernel is  $W \cap V_+$ ,  $F=+1, \varepsilon=+1$ ; cokernel  $V_- \cap W^\perp$ ,  $F=-1, \varepsilon=-1$ , get the +1 eigenspace for g.

~~Bottomless~~ Tree ??



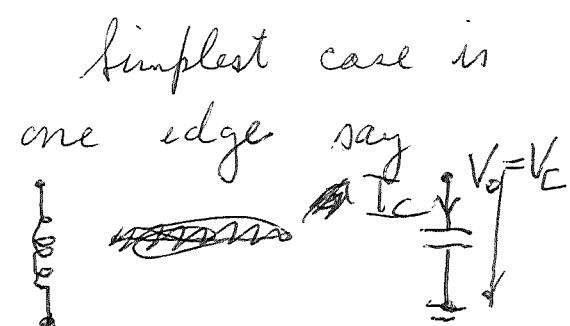
Take the ~~case~~ first

$$\{V_0\} = \bar{C}^0 \xrightarrow{\sim} C_c^1 = \{V_c\}.$$

$$Cs(V_c) = V_0$$

$$I_0 = I_c = CsV_0$$

This is the case where  $H^1 = 0 = H_1$ ,

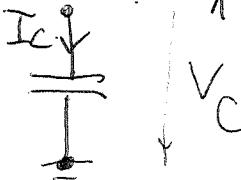


~~constraints~~  $I_c = 0$

$$\text{IVP: } CV_c = I_c$$

$$Cs\hat{V}_c - V_c(0) = \hat{I}_c$$

$$Cs\hat{V}_c - \hat{I}_c = CV_c(0)$$

$\phi'$  circuit  Go over the 4<sup>th</sup> cases. 1st  $\bar{C}^o \rightarrow C'_C$  constraint Kirchhoff I  $\therefore I_C = 0$ .  $C \dot{V}_C = I_C$

L.T.  $C_s \hat{V}_C = -CV_C(0) + \hat{I}_C$

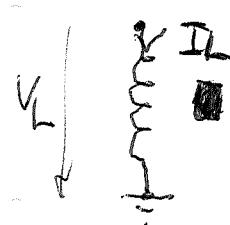
$C_s \hat{V}_C = -CV_C(0)$

$\hat{V}_C = \frac{1}{s} V_C(0)$

$V_C(t) = H(t) V_C(0)$ .

$$\int_0^\infty e^{-st} \hat{F}(t) dt = \int_0^\infty ((e^{-st}\hat{F})' + se^{-st}\hat{F}) dt$$

$$\hat{F} = -F(0) + s\hat{F}$$



$I_L = 0$ . Kirchhoff I

$L \dot{I}_L = V_L$  ??

~~constraint~~  $L \dot{I}_L = L_s \hat{I}_L - L I_L(0)$

$\bar{C}^o \sim C'_L$

$\{V^o\} \longrightarrow \{V_L\}$

$\bar{C}_o \leftarrow \sim C_{o,L}$

$\{I_L\}$

apparently this ~~circuit~~ circuit has no motion

$0 = \bar{C}^o \sim C'_L \sim H^I$

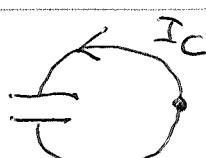


$V_L = 0$



$0 \sim C_{o,L} \sim H_I$

$0 = \bar{C}^o \sim C'_C \sim H^I$



$V_C = 0$

$C_{o,C} \sim H_I$

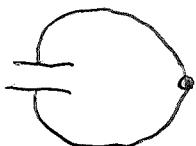
$C \dot{V}_C = I_C$

no motion at all

$X'$  You've examined four cases, and found 2 with no motion. These are



$$\tilde{C}^0 \cong C_L^1 \quad \varepsilon = -1, F = 1$$



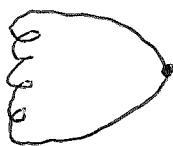
$$C_C^1 \cong H^1 \quad \varepsilon = +1, F = -1$$

Other two which have motion

These give the  $g = -1$  eigenspace



$$\tilde{C}^0 \cong C_C^1 \quad \varepsilon = +1, F = +1$$



$$C_{L,C} \cong H_1 \quad \varepsilon = -1, F = -1$$

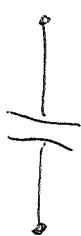
These give the  $g = +1$  eigenspace.

Tree

$$\begin{array}{c} \tilde{C}^0 \cong C^1 \\ \uparrow T_s \\ \tilde{C}_0 \cong C_1 \end{array}$$

What happens to the IVP?

Is it true that  $T_s$  is a comp. to the Kirchhoff space  $C^1$ .



$$\begin{array}{c} \tilde{C}^0 \cong C_C^1 \\ \downarrow s \\ \tilde{C}_0 \cong C_{L,C} \end{array}$$

$$C_C^1 = \{V_C\}$$

$$C_{L,C} = \{I_C\}$$

$$\dot{V}_C = I_C$$

$$\hat{I}_C = \hat{V}_C = s\hat{V}_C - V_C(0)$$

$$T_s = \left\{ \begin{pmatrix} \hat{V}_C \\ \hat{I}_C \end{pmatrix} \mid s\hat{V}_C = \hat{I}_C \right\}$$

$$\tilde{C}^0 = \left\{ \begin{pmatrix} V_C \\ 0 \end{pmatrix} \right\}$$

$$\mathcal{C}^o \Rightarrow C_L^I \quad C_L^I = \{V_L\}$$

$$\bar{\mathcal{C}}_o \Leftarrow C_{I,L} \quad C_{I,L} = \{I_L\}$$


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$$\dot{I}_L = V_L \quad \hat{V}_L = \hat{\dot{I}}_L = s\dot{I}_L - I_L(0)$$

$$\bar{\mathcal{C}}^o = \begin{pmatrix} C_C^I \\ C_L^I \end{pmatrix}$$

~~Defn~~

$$\bar{\mathcal{C}}_o = \begin{pmatrix} C_{I,C} \\ C_{I,L} \end{pmatrix}$$

$$\begin{array}{l} \{V_C\} \\ \{V_L\} \\ \{I_C\} \\ \{I_L\} \end{array}$$


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~~Defn~~ You probably need to keep track of  $s=0,\infty$ .

Consider a tree with L and C type edges. You want the Impedance correspondence  $\Gamma_s \subset \mathcal{C}^o = \begin{pmatrix} C^I \\ C_I \end{pmatrix}$  to be complementary to the Kirchhoff space  $\begin{pmatrix} C^I \\ 0 \end{pmatrix}$

$$\begin{pmatrix} C^I \\ C_I \end{pmatrix} = \begin{pmatrix} C_C^I \\ C_{I,C} \end{pmatrix} \oplus \begin{pmatrix} C_L^I \\ C_{I,L} \end{pmatrix}$$

||                           ||

$$\Gamma_s = \begin{pmatrix} 1 \\ s \end{pmatrix} \boxed{C_C^I} \oplus \begin{pmatrix} s \\ 1 \end{pmatrix} \boxed{C_{I,L}}$$

when  $\Gamma_s \cap \begin{pmatrix} C^I \\ 0 \end{pmatrix} \neq 0$ . If  $s=\infty$ , this intersection is  $C_L^I$ . If  $s=0$ , this intersection is  $C_C^I$ .

$\omega'$  function of a graph  $\stackrel{?}{=}$  characteristic polynomial of some correspondence?  $\oplus$  Kronecker module. something related to fixpts of iterates. There are interesting examples related to geodesic flows.

General  $W \subset \begin{pmatrix} V_+ \\ V_- \end{pmatrix}$ . You want the projection  $W \rightarrow V_+$  to be an isom. so that  $W = \begin{pmatrix} 1 \\ T \end{pmatrix} V_+$   $T: V_+ \rightarrow V_-$  So look at the kernel of  $W \rightarrow V_+$  i.e.  $W \cap V_-$ , where  $F=1$  and  $\varepsilon=-1$ ,  $\therefore g=-1$ . Split off the kernel, look at image  $W \hookrightarrow V_+$ , orthog space is  $W^\perp \cap V_+$ , where  $F=-1, \varepsilon=+1$  so  $g=+1$ .

Now you want to find a picture. First make  $T$  invertible. Consider the projection  $W \rightarrow V_-$  remove the kernel  $W \cap V_+$ , and remove the "cokernel"  $V_- \cap W$ . These two spaces have  $F=-1, \varepsilon=+1$  and  $F=+1, \varepsilon=-1$  resp., hence comprise all of  $g \neq +1$  eigenspace.

You would like a picture consisting of the "core" where  $W$  projects bijectively to both  $V_+$  and  $V_-$  (i.e.  $T: V_+ \rightarrow V_-$  is invertible), and 4 "wings". Where to start? The case  $W = \begin{pmatrix} 1 \\ T \end{pmatrix} V_+$  where  $T$  is not invertible.

Let's start again, choose  $V_+, V_-$  with pos. def. scalar products, put  $V = \begin{pmatrix} V_+ \\ V_- \end{pmatrix}$  and let  $W \subset \begin{pmatrix} V_+ \\ V_- \end{pmatrix}$ . Let's digress

a little back to  $W$  a linear retract of  $\begin{pmatrix} V_+ \\ V_- \end{pmatrix}$  pd. fac.

$$W \xleftarrow{\alpha^* \alpha^*} \begin{pmatrix} V_+ \\ V_- \end{pmatrix} \xleftarrow{\alpha^*} W \quad \underbrace{\alpha^*_{+} d_+}_{h_+} + \underbrace{\alpha^*_{-} d_-}_{h_-} = 1_W$$

$$h_+ = \frac{1}{1+\omega^2} \quad h_- = \frac{\omega^2}{1+\omega^2} \quad \frac{s + \omega^2 s^{-1}}{1+\omega^2} = \frac{s^2 + \omega^2}{s(1+\omega^2)}$$

$$\begin{pmatrix} s & T^* \\ -T & s \end{pmatrix} = \overbrace{s \begin{pmatrix} 1 & 0 \\ 0 & 1 \end{pmatrix} - \begin{pmatrix} 0 & -T \\ T & 0 \end{pmatrix}}^{(1-\omega^2)(1+\omega^2)} \overbrace{(1+\omega^2)}^{s^2 + \omega^2}$$